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**18ECL48: ANALOG CIRCUITS LAB.....2**

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## A. LABORATORY INFORMATION

### 1. Lab Overview

Degree:	B.E	Program:	EC
Year / Semester :	2 / 4	Academic Year:	2019-20
Course Title:	Analog Circuits Lab	Course Code:	18ECL48
Credit / L-T-P:	2/ 2-0-0	SEE Duration:	180 Minutes
Total Contact Hours:	36 Hrs	SEE Marks:	100 Marks
CIA Marks:	40	Assignment	- -----
Course Plan Author:	Arun Kumar R	Sign	Dt :
Checked By:		Sign	Dt :

### 2. Lab Content

Unit	Title of the Experiments	Lab Hours	Concept	Blooms Level
1	Design and setup the Common Source JFET/MOSFET amplifier and plot the frequency response	3	Frequency response of JFET/MOSFET	L3
2	Design and set up the BJT common emitter voltage amplifier with and without feedback and determine the gain- bandwidth product, input and output impedances.	3	BJT common emitter amplification	L3
3	Design and set-up BJT/FET i) Colpitts Oscillator, and ii) Crystal Oscillator	3	Oscillator	L3
4	Design active second order Butterworth low pass and high pass filters.	3	Filters	L3
5	Design Adder, Integrator and Differentiator circuits using Op-Amp	3	Opamp applications	L3
6	Test a comparator circuit and design a Schmitt trigger for the given UTP and LTP values and obtain the hysteresis.	3	comparator	L3
7	Design 4 bit R – 2R Op-Amp Digital to Analog Converter (i) using 4 bit binary input from toggle switches and (ii) by generating digital inputs using mod-16 counter.	3	DAC	L3
8	Design Monostable and a stable Multivibrator using 555 Timer.	3	Multivibrator	L3
9	Simulation using EDA software- RC Phase shift oscillator and Hartley oscillator	3	Oscillator	L3
10	Simulation using EDA software- Narrow Band-pass Filter and Narrow band-reject filter	3	Filter	L3
11	Simulation using EDA software- Precision Half and full wave rectifier	3	Rectifier	L3
12	Simulation using EDA software-Monostable and A stable Multivibrator using 555 Timer.	3	Multivibrator	L3

### 3. Lab Material

Unit	Details	Available
1	Text books	
	Linear Integrated CircuitsII, D. Roy Choudhury and Shail B. Jain,4thedition, Reprint 2006, New Age International ISBN 978-81-224-3098-1.	
	Operational Amplifiers and Linear IC'sII, David A. Bell, 2nd edition, PHI/Pearson,2004. ISBN 978-81-203-2359-9	
	David A Bell, "Fundamentals of Electronic Devices and Circuits Lab Manual, 5th Edition, 2009, Oxford University Press.	
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2	Reference books	
i	Ramakant A Gayakwad, –Op-Amps and Linear Integrated Circuits, Pearson, 4th Ed, 2015. ISBN 81-7808-501-1.	In Lib and dept
ii	Robert L. Boylestad and Louis Nashelsky, “Electronics devices and Circuit theory”, Pearson, 10th Edition, 2012, ISBN: 978-81-317-6459-6.	In Lib
iii	K. A. Navas, “Electronics Lab Manual”, Volume I, PHI, 5th Edition, 2015, ISBN:9788120351424.	In Lib
iv	B Somanathan Nair, –Linear Integrated Circuits: Analysis, Design & Applications, Wiley India, 1st Edition, 2015	In Lib
v	James Cox, –Linear Electronics Circuits and Devices II, Cengage Learning, Indian Edition, 2008, ISBN-13: 978-07-668-3018-7.	In Lib
3	Others (Web, Video, Simulation, Notes etc.)	In Lib
i	Data Sheet: <a href="http://www.ti.com/lit/ds/symlink/tl081.pdf">http://www.ti.com/lit/ds/symlink/tl081.pdf</a>	
	<ul style="list-style-type: none"> <li><a href="https://www.youtube.com/watch?v=CoOOm3NEMfg">https://www.youtube.com/watch?v=CoOOm3NEMfg</a></li> <li><a href="https://www.youtube.com/watch?v=6A8otDArahM">https://www.youtube.com/watch?v=6A8otDArahM</a></li> <li><a href="https://www.youtube.com/watch?v=1fgw-ONIAcc">https://www.youtube.com/watch?v=1fgw-ONIAcc</a></li> <li><a href="https://www.youtube.com/watch?v=YzckQWwkzWs">https://www.youtube.com/watch?v=YzckQWwkzWs</a></li> <li><a href="https://www.youtube.com/watch?v=lc6QT8VjqVc">https://www.youtube.com/watch?v=lc6QT8VjqVc</a></li> <li><a href="https://www.youtube.com/watch?v=sKnLBWA6UdE">https://www.youtube.com/watch?v=sKnLBWA6UdE</a></li> <li><a href="https://www.youtube.com/watch?v=BCjnYMNCKGc">https://www.youtube.com/watch?v=BCjnYMNCKGc</a></li> <li><a href="https://www.youtube.com/watch?v=5-ohKRWeod4">https://www.youtube.com/watch?v=5-ohKRWeod4</a></li> <li><a href="https://www.youtube.com/watch?v=Pc1aFloxSMw">https://www.youtube.com/watch?v=Pc1aFloxSMw</a></li> <li><a href="https://www.youtube.com/watch?v=XES0QUi8ttY">https://www.youtube.com/watch?v=XES0QUi8ttY</a></li> <li><a href="https://www.youtube.com/watch?v=ypV6gdIJJU4">https://www.youtube.com/watch?v=ypV6gdIJJU4</a></li> <li><a href="https://www.youtube.com/watch?v=iJYm_BGqa1A">https://www.youtube.com/watch?v=iJYm_BGqa1A</a></li> <li><a href="https://www.youtube.com/watch?v=k3XgLk2H1w8">https://www.youtube.com/watch?v=k3XgLk2H1w8</a></li> <li><a href="https://www.youtube.com/watch?v=GH-JFXbOcZg">https://www.youtube.com/watch?v=GH-JFXbOcZg</a></li> <li><a href="https://www.youtube.com/watch?v=v9sSRF76DDU">https://www.youtube.com/watch?v=v9sSRF76DDU</a></li> <li><a href="https://www.youtube.com/watch?v=ZuFKU9O8FTs">https://www.youtube.com/watch?v=ZuFKU9O8FTs</a></li> <li><a href="https://www.youtube.com/watch?v=lh768hHRsxg">https://www.youtube.com/watch?v=lh768hHRsxg</a></li> <li><a href="https://www.youtube.com/watch?v=7jGobEEyD7w">https://www.youtube.com/watch?v=7jGobEEyD7w</a></li> </ul>	
	Nptel.ac.in for videos	

#### 4. Lab Prerequisites:

SNo	Course Code	Base Course: Course Name	Topic / Description	Sem	Remarks
1	18ELN14/24	Basic Electronics	OPAMP / Rectifiers / BJT & FET working operation	1/2	

Note: If prerequisites are not taught earlier, GAP in curriculum needs to be addressed. Include in Remarks and implement in B.5.

#### 5. General Instructions

SNo	Instructions	Remarks
1	Observation book and Lab record are compulsory.	
2	Students should report to the concerned lab as per the time table.	
3	After completion of the program, certification of the concerned staff in-charge in	

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	the observation book is necessary.	
4	Student should bring a notebook of 100 pages and should enter the readings /observations into the notebook while performing the experiment.	
5	The record of observations along with the detailed experimental procedure of the experiment in the Immediate last session should be submitted and certified staff member in-charge.	
6	Should attempt all problems / assignments given in the list session wise.	
7	When the experiment is completed, should disconnect the setup made by them, and should return all the components/instruments taken for the purpose.	
8	Any damage of the equipment or burn-out components will be viewed seriously either by putting penalty or by dismissing the total group of students from the lab for the semester/year	
9	Completed lab assignments should be submitted in the form of a Lab Record in which you have to write the Components required, Theory, Procedure, tabular column and output for various inputs given	

## 6. Lab Specific Instructions

SNo	Specific Instructions	Remarks
1	Rigup the circuits as shown in the lab manual for each experiment	
2	Turn on the supply and Apply the proper input when it is required	
3	Observe the output , note down the readings and compare it with theoretical value	
4	Plot the graph using graph/ semilog sheets	
5	Turn off the supply & Disconnect the circuit	

## B. OBE PARAMETERS

### 1. Lab / Course Outcomes

#	COs	Teach. Hours	Concept	Instr Method	Assessment Method	Blooms ' Level
1	Design the circuits using BJT and FET	12	BJT / FET performance characteristics	Lecture and Demonstrate	Test and Viva	L3
2	Design the circuits using Op-Amp	12	OPAMPs for different applications	Lecture and Demonstrate	Test and Viva	L3
3	Simulate and analyze analog circuits that uses transistor and ICs for different electronic applications.	12	555 timer and Simulation	Lecture and Demonstrate	Test and Viva	L3

Note: Identify a max of 2 Concepts per unit. Write 1 CO per concept.

### 2. Lab Applications

SNo	Application Area	CO	Level
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1	frequency-determining component, a wafer of quartz crystal or ceramic with electrodes connected to it.	CO1	L3
2	Used in speech processing for communications (radio) applications & audio compression.	CO1	L3
3	Used in RF, radio, small power supply units, medical field etc	CO1	L3
4	In some Capacitor to power the high-side NMOS driver.	CO1	L3
5	Apply voltage divider bias to find gain bandwidth & frequency response	CO1	L3
6	Used in buffer amplifier, electronic switch, phase shift oscillator etc	CO1	L3
7	widely used for switching and amplifying electronic signals in the electronic devices.	CO1	L3
8	Oscillators is used in microprocessor for clock circuits	CO1	L3
9	used in low-cost design devices and mobile devices	CO2	L3
10	Differentiator is used in signal amplifier	CO2	L3
11	Integrator used in wave shaping circuit	CO2	L3
12	Active filters used in demodulator circuit	CO2	L3
13	R-2R ladder network used in DAC	CO2	L3
14	Multivibrators are used for time delay calculation	CO3	L3
15	Used as external triggers	CO3	L3

Note: Write 1 or 2 applications per CO.

### 3. Articulation Matrix

#### (CO - PO MAPPING)

#	Course Outcomes COs	Program Outcomes												Level	
		PO 1	PO 2	PO 3	PO 4	PO 5	PO 6	PO 7	PO 8	PO 9	PO 10	PO 11	PO 12		
18EC48.1	Design the circuits using BJT and FET	3	3	2							2	2			L3
18EC48.2	Design the circuits using Op-Amp and 555 timer for different applications	3	3	2							2	2			L3
18EC48.3	Simulate and analyze analog circuits that uses transistor and ICs for different electronic applications.	3	3	2		2					2	2			L3

Note: Mention the mapping strength as 1, 2, or 3

### 4. Curricular Gap and Content

SNo	Gap Topic	Actions Planned	Schedule Planned	Resources Person	PO Mapping
1					
2					
3					
4					
5					

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Note: Write Gap topics from A.4 and add others also.

Note: Anything not covered above is included here.

## C. COURSE ASSESSMENT

### 1. Course Coverage

Unit	Title	Teaching Hours	No. of question in Exam							CO	Levels
			CIA-1	CIA-2	CIA-3	Asg-1	Asg-2	Asg-3	SEE		
1	Design and setup the Common Source JFET/MOSFET amplifier and plot the frequency response	03	1	-	-	-	-	-	1	CO1	L4
2	Design and set up the BJT common emitter voltage amplifier with and without feedback and determine the gain- bandwidth product, input and output impedances.	03	1	-	-	-	-	-	1	CO1	L4
3	Design and set-up BJT/FET i) Colpitts Oscillator, and ii) Crystal Oscillator	03	2	-	-	-	-	-	2	CO1	L4
4	Design active second order Butterworth low pass and high pass filters.	03	2	-	-	-	-	-	2	CO2	L4
5	Design Adder, Integrator and Differentiator circuits using Op-Amp	03	2	-	-	-	-	-	2	CO2	L4
6	Test a comparator circuit and design a Schmitt trigger for the given UTP and LTP values and obtain the hysteresis.	03	2	-	-	-	-	-	2	CO2	L4
8	Design Monostable and a stable Multivibrator using 555 Timer.	03	-	2	-	-	-	-	2	CO2	L4
9	Simulation using EDA software- RC Phase shift oscillator and Hartley oscillator	03	-	2	-	-	-	-	2	CO2	L4
10	Simulation using EDA software- Narrow Band-pass Filter and Narrow band-reject filter	03	-	2	-	-	-	-	2	CO3	L4
11	Simulation using EDA software- Precision Half and full wave rectifier	03	-	2	-	-	-	-	2	CO3	L4
12	Simulation using EDA software- Monostable and A stable Multivibrator using 555 Timer.	03	-	2	-	-	-	-	2	CO3	L4
-	<b>Total</b>	<b>36</b>	<b>10</b>	<b>11</b>	-	-	-	-	<b>21</b>	-	-

Note: Write CO based on the theory course.

### 2. Continuous Internal Assessment (CIA)

Evaluation	Weightage in Marks	CO	Levels
CIA Exam – 1	30	CO1, CO2	L2
CIA Exam – 2	30	CO3, CO4	L4
CIA Exam – 3	30	CO4	L4

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Assignment - 1	05	CO1, CO2	L2
Assignment - 2	05	CO3,C04	L4
Assignment - 3	05	CO4	L4
Other Activities – define – Slip test		CO1 to CO4	L2, L3, L4 . . .
<b>Final CIA Marks</b>	<b>40</b>	-	-

SNo	Description	Marks
1	Observation and Weekly Laboratory Activities	05 Marks
2	Record Writing	10 Marks for each Expt
3	Internal Exam Assessment	20 Marks
4	Internal Assessment	40 Marks
5	SEE	60Marks
-	<b>Total</b>	<b>100 Marks</b>

## D. EXPERIMENTS

### Experiment 01 : COMMON SOURCE JFET/MOSFET AMPLIFIER

-	Experiment No.:	1	Marks	10	Date Planned	Date Conducted																																								
1	Title	<b>COMMON SOURCE JFET/MOSFET AMPLIFIER</b>																																												
2	Course Outcomes	Design analog circuits using BJT/FETs and evaluate their performance characteristics.																																												
3	Aim	To design, setup and plot the frequency response of Common Source JFET amplifier and obtain the bandwidth.																																												
4	Material / Equipment Required	<table border="1"> <thead> <tr> <th>Sl. No.</th> <th>Particulars</th> <th>Range</th> <th>Quantity</th> </tr> </thead> <tbody> <tr> <td>1.</td> <td>JFET (BFW10)</td> <td>-</td> <td>1</td> </tr> <tr> <td>2.</td> <td>Capacitor</td> <td>47<math>\mu</math>F,0.1<math>\mu</math>F</td> <td>1,2</td> </tr> <tr> <td>3.</td> <td>Resistor</td> <td>2M<math>\Omega</math>, 820<math>\Omega</math>,330<math>\Omega</math></td> <td>1,1</td> </tr> <tr> <td>4.</td> <td>Power Supply</td> <td>0-30V</td> <td>1</td> </tr> <tr> <td>5.</td> <td>Signal Generator</td> <td>2MHz</td> <td>1</td> </tr> <tr> <td>6.</td> <td>CRO</td> <td>-</td> <td>1</td> </tr> <tr> <td>7.</td> <td>Multimeter</td> <td>-</td> <td>1</td> </tr> <tr> <td>8.</td> <td>Spring Board</td> <td>-</td> <td>1</td> </tr> <tr> <td>9.</td> <td>Connecting Wires</td> <td>-</td> <td>-</td> </tr> </tbody> </table>					Sl. No.	Particulars	Range	Quantity	1.	JFET (BFW10)	-	1	2.	Capacitor	47 $\mu$ F,0.1 $\mu$ F	1,2	3.	Resistor	2M $\Omega$ , 820 $\Omega$ ,330 $\Omega$	1,1	4.	Power Supply	0-30V	1	5.	Signal Generator	2MHz	1	6.	CRO	-	1	7.	Multimeter	-	1	8.	Spring Board	-	1	9.	Connecting Wires	-	-
Sl. No.	Particulars	Range	Quantity																																											
1.	JFET (BFW10)	-	1																																											
2.	Capacitor	47 $\mu$ F,0.1 $\mu$ F	1,2																																											
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4.	Power Supply	0-30V	1																																											
5.	Signal Generator	2MHz	1																																											
6.	CRO	-	1																																											
7.	Multimeter	-	1																																											
8.	Spring Board	-	1																																											
9.	Connecting Wires	-	-																																											
5	Theory, Formula, Principle, Concept	The JFET gate voltage $V_g$ is biased through the potential divider network set up by the resistors $R_1$ and $R_2$ and is biased to operate within its saturation region which is equivalent to the active region of the BJT.																																												

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		<p>Unlike the BJT, the junction FET takes virtually no input gate current allowing the gate to be treated as an open circuit. Then no input characteristics curves are required.</p> <p>Since the n-channel JFET is a depletion mode device, a negative gate voltage with respect to the source is required to modulate or control the drain current. This negative voltage can be provided by biasing from a separate power supply voltage or by self biasing arrangements as long as steady current flows through the JFET even when there is no input signal present and <math>V_g</math> maintains a reverse bias of the gate source p-n junction.</p> <p>The input signal of the common source JFET amplifier is applied between the gate terminals with a constant value of gate voltage applied. The JFET operates within its ohmic region acting like a linear resistive device. The drain circuit contains the load resistor <math>R_D</math>. The output voltage is developed across this load resistance.</p>
6	Procedure, Program, Activity, Algorithm, Pseudo Code	<ul style="list-style-type: none"> <li>•Check all the components and equipments for their good working condition.</li> <li>•Connections are made as shown in the circuit diagram.             <ul style="list-style-type: none"> <li>•By keeping the voltage knobs in minimum position and current knob in maximum position switch on the power supply.</li> </ul> </li> <li>•By disconnecting the AC source measure the quiescent point.</li> <li>•<b>To find frequency response:</b> <ul style="list-style-type: none"> <li>•Connect the AC source. Keeping the frequency of the Ac source in mid band region (say 10 kHz) adjust the amplitude to get the distortion less output. Note down the amplitude of the input signal.</li> <li>•Keeping the input amplitude constant, Vary the frequency in suitable steps and note down the corresponding output amplitude.</li> <li>•Calculate <math>A_v</math> and gain in decibels. Plot a graph of frequency <math>V_s</math> gain in dB. From the graph calculate <math>f_L, f_H</math> and band width.</li> </ul> </li> <li>•Calculate figure of merit.</li> <li>•<b>To find the input impedance (<math>Z_i</math>):</b></li> </ul>





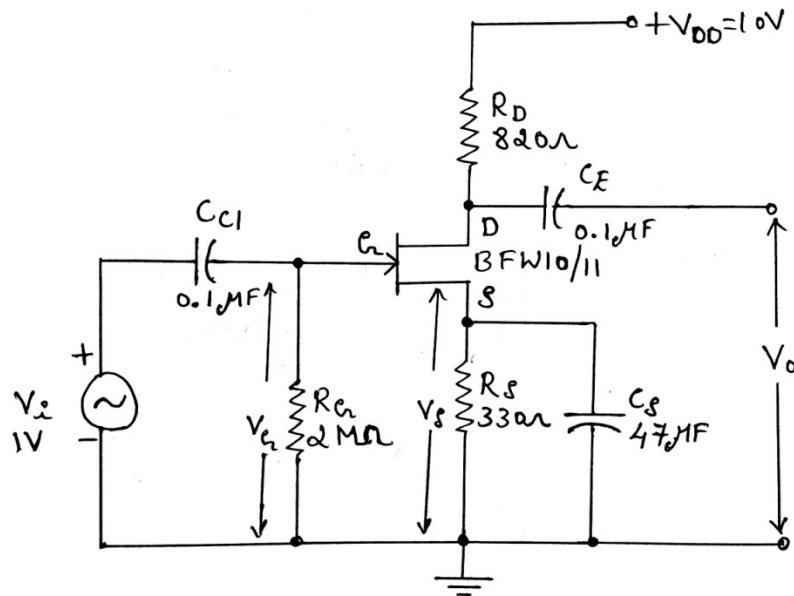
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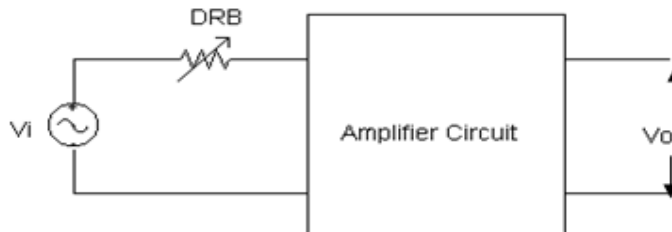
- Connections are made as shown in the diagram.
- Keeping the DRB in its minimum position, apply input signal at mid band frequency (say 10kHz) and adjust the amplitude of the input signal to get distortion less output. Note down the output amplitude.
- Vary the DRB until the output amplitude becomes half of its previous value. The corresponding DRB value gives the input impedance.
- To find the output impedance ( $Z_o$ ):**
- Connections are made as shown in the diagram.
- Keeping the DRB in its maximum position, apply input signal at mid band frequency (say 10k Hz) and adjust the amplitude of the input signal to get distortion less output. Note down the output amplitude.

Vary the DRB until the output amplitude becomes half of its previous value. The corresponding DRB value gives the output impedance.

7 Block, Circuit, Model Diagram, Reaction Equation, Expected Graph



**Circuit to find input impedance ( $Z_i$ ):**

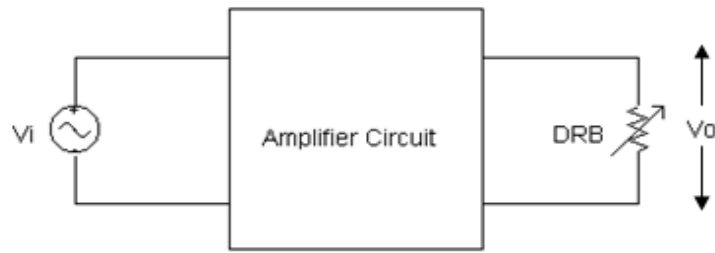


**Circuit to find output impedance ( $Z_o$ ):**



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8	Observation Look-up Output	Table,					
		Table,					
		<b>Frequency in Hz</b>	$V_o$ in Volt	$A_v = \frac{V_o}{V_i}$	<b>Gain in dB=20 log</b>	<b>log</b>	
					$A_v$		

9	Sample Calculations	<p>Given <math>V_{DD} = 10V</math>, <math>V_{GS(off)} = -4V</math>, <math>I_{DSS(max)} = 12mA</math>, <math>R_G = 2M\Omega</math></p> <p>Formulae,</p> $I_D = I_{DSS} \left( \frac{1 - V_{GS}}{V_{GS(off)}} \right)^2 \text{ -----(1)}$ <p>When <math>V_G = 0</math>, Then <math>V_S = -V_{GS}</math></p> <p>But <math>V_S = I_D \times R_S</math></p> <p>When <math>V_G = 0</math>, <math>I_D = I_{DSS}</math></p> $V_S = I_{DSS} \times R_S$ $I_{DSS} \times R_S = -V_{GS(off)}$ $R_S = \frac{-(-4)}{12mA} = 333\Omega$ <p>Choose <math>R_S = 330\Omega</math></p> <p>From Eq. (1)</p> $I_D = I_{DSS} \left( \frac{1 - I_D R_S}{V_{GS(off)}} \right)^2$ $I_D = I_{DSS} \left( \frac{1 + I_D^2 \times R_S^2 - I_D R_S}{16} \right)$
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		$I_D = 12 \times 10^{-3} \left( \frac{1 + I_D^2 \times 330^2}{16} - \frac{I_D \times 330}{2} \right)$ $81.675 I_D^2 - 2.98 I_D + 12 \times 10^{-3} = 0$ $I_D = 4.6 \text{ mA} \quad \text{or} \quad I_D = 31.9 \text{ mA}$ <p>Since <math>I_D</math> cannot be greater than <math>I_{DSS}</math>, Choose <math>I_D = 4.6 \text{ mA}</math></p> <p>Assume <math>V_{DS} = 50\% V_{DD}</math>, <math>V_{DS} = 5 \text{ V}</math></p> <p>Applying KVL to output circuit</p> $V_{DD} = I_D R_D + V_{DS} + I_D R_S$ $V_{DD} - V_{DS} = I_D (R_D + R_S)$ $\frac{10 \text{ V} - 5 \text{ V}}{I_D} = (R_D + R_S)$ $\frac{5 \text{ V}}{4.6 \text{ mA}} = (R_D + R_S)$ $R_D = \frac{5 \text{ V}}{4.6 \text{ mA}} - 330 \Omega$ $R_D = 756 \Omega$ <p>Choose <math>R_D = 820 \Omega</math></p>
--	--	--

10	Graphs, Outputs	<p style="text-align: center;"><math>f_L = \text{Lower cutoff frequency} \quad f_H = \text{Higher cutoff frequency}</math></p>
----	-----------------	--

11	Results & Analysis	<ol style="list-style-type: none"> <li>1. Quiescent point: <math>V_{DS} = \underline{\hspace{2cm}}</math> V and <math>I_D = \underline{\hspace{2cm}}</math> mA.</li> <li>2. Voltage Gain (<math>A_V</math>) = <math>\underline{\hspace{2cm}}</math> ( in mid band region).</li> <li>3. Bandwidth (BW) = <math>\underline{\hspace{2cm}}</math> Hz.</li> <li>4. figure of merit (FM) = <math>\underline{\hspace{2cm}}</math> Hz.</li> <li>5. Input impedance (<math>Z_i</math>) = <math>\underline{\hspace{2cm}}</math> <math>\Omega</math>, Output Impedance (<math>Z_o</math>) = <math>\underline{\hspace{2cm}}</math> <math>\Omega</math>.</li> </ol>
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		=
12	Application Areas	Used in CRO used in electronic voltmeter
13	Remarks	
14	Faculty Signature with Date	

## Experiment 02 : BJT COMMON EMITTER AMPLIFIER

-	Experiment No.:	2	Marks	10	Date Planned	Date Conducted																																								
1	Title	<b>BJT COMMON EMITTER AMPLIFIER</b>																																												
2	Course Outcomes	Design analog circuits using BJT/FETs and evaluate their performance characteristics.																																												
3	Aim	To design and set up the BJT common emitter amplifier using voltage divider bias with and without feedback and determine the gain- bandwidth product from its frequency response.																																												
4	Material Equipment Required	<table border="1" style="width: 100%; border-collapse: collapse;"> <thead> <tr> <th>Sl. No.</th> <th>Particulars</th> <th>Range</th> <th>Quantity</th> </tr> </thead> <tbody> <tr> <td>1.</td> <td>Transistor (SL100/CL100)</td> <td>-</td> <td>1</td> </tr> <tr> <td>2.</td> <td>Resistor</td> <td>47KΩ, 1.8KΩ, 8.2KΩ, 470Ω</td> <td>1,1,1,1</td> </tr> <tr> <td>3.</td> <td>Capacitor</td> <td>0.47μF, 47μF</td> <td>2,1</td> </tr> <tr> <td>4.</td> <td>Signal Generator</td> <td>20MHz</td> <td>1</td> </tr> <tr> <td>5.</td> <td>CRO</td> <td>-</td> <td>1</td> </tr> <tr> <td>6.</td> <td>Power Supply</td> <td>0-30V</td> <td>1</td> </tr> <tr> <td>7.</td> <td>Multimeter</td> <td>-</td> <td>1</td> </tr> <tr> <td>8.</td> <td>Spring Board</td> <td>-</td> <td>1</td> </tr> <tr> <td>9.</td> <td>Connecting Wires</td> <td>-</td> <td>-</td> </tr> </tbody> </table>					Sl. No.	Particulars	Range	Quantity	1.	Transistor (SL100/CL100)	-	1	2.	Resistor	47KΩ, 1.8KΩ, 8.2KΩ, 470Ω	1,1,1,1	3.	Capacitor	0.47μF, 47μF	2,1	4.	Signal Generator	20MHz	1	5.	CRO	-	1	6.	Power Supply	0-30V	1	7.	Multimeter	-	1	8.	Spring Board	-	1	9.	Connecting Wires	-	-
Sl. No.	Particulars	Range	Quantity																																											
1.	Transistor (SL100/CL100)	-	1																																											
2.	Resistor	47KΩ, 1.8KΩ, 8.2KΩ, 470Ω	1,1,1,1																																											
3.	Capacitor	0.47μF, 47μF	2,1																																											
4.	Signal Generator	20MHz	1																																											
5.	CRO	-	1																																											
6.	Power Supply	0-30V	1																																											
7.	Multimeter	-	1																																											
8.	Spring Board	-	1																																											
9.	Connecting Wires	-	-																																											
5	Theory, Formula, Principle, Concept	<p>The single staged common emitter amplifier circuit uses voltage divider biasing. This type of biasing arrangement uses two resistors as a potential divider network across the supply with their centre point supplying the required base bias voltage to the transistor. Voltage divider biasing is commonly used in the design of bipolar transistor amplifier circuits.</p> <p>In common emitter amplifier circuits, capacitors <math>C_1</math> and <math>C_2</math> are used as coupling capacitors to separate the AC signals from the DC biasing voltage. This ensures that the biasing condition set up for the circuit to operate correctly is not affected by any additional amplifier stages as the capacitors will only pass AC signals and block any DC component. The output AC signal is then superimposed on the biasing of the following stages. Also a bypass capacitor <math>C_E</math> is included in the emitter leg circuit.</p>																																												



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6	Procedure, Program, Activity, Algorithm, Pseudo Code	<p>Check all the components and equipments for their good working condition. Connections are made as shown in the circuit diagram. By keeping the voltage knobs in minimum position and current knob in maximum position switch on the power supply. By disconnecting the AC source measure the quiescent point.</p> <p><b>To find frequency response:</b></p> <ol style="list-style-type: none"><li>1. Connect the AC source. Keeping the frequency of the Ac source in mid band region (say 10 kHz) adjust the amplitude to get the distortion less output. Note down the amplitude of the input signal.</li><li>2. Keeping the input amplitude constant, Vary the frequency in suitable steps and note down the corresponding output amplitude.</li><li>3. Calculate <math>A_v</math> and gain in decibels. Plot a graph of frequency Vs gain in dB. From the graph calculate <math>f_L, f_H</math> and band width.</li><li>4. Calculate figure of merit.</li></ol> <p><b>To find the input impedance (<math>Z_i</math>):</b></p> <ol style="list-style-type: none"><li>1. Connections are made as shown in the diagram.</li><li>2. Keeping the DRB in its minimum position, apply input signal at mid band frequency (say 10kHz) and adjust the amplitude of the input signal to get distortion less output. Note down the output amplitude.</li><li>3. Vary the DRB until the output amplitude becomes half of its previous value. The corresponding DRB value gives the input impedance.</li></ol> <p><b>To find the output impedance (<math>Z_o</math>):</b></p> <ol style="list-style-type: none"><li>1. Connections are made as shown in the diagram.</li><li>2. Keeping the DRB in its maximum position, apply input signal at mid band frequency (say 10k Hz) and adjust the amplitude of the input signal to get distortion less output. Note down the output amplitude.</li><li>3. Vary the DRB until the output amplitude becomes half of its previous value. The corresponding DRB value gives the output impedance.</li></ol> <p><b>Result:</b></p>
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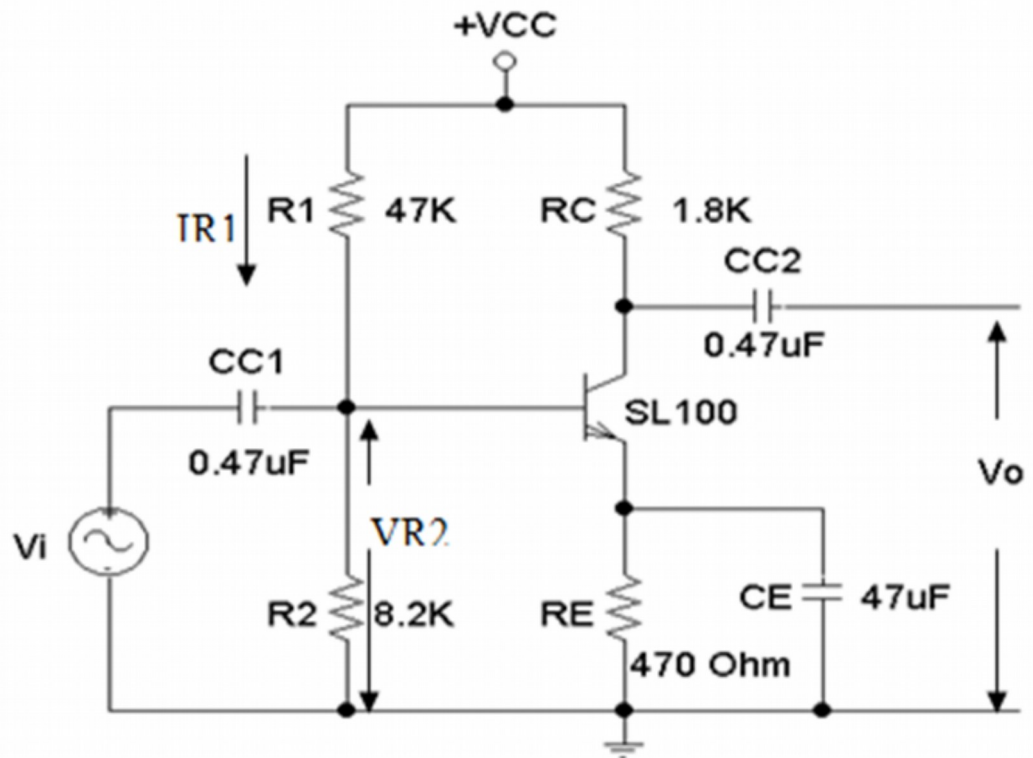
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1. Quiescent point:  $V_{CE} = 5V$ ,  $I_{C2} = 2mA$ .
2. Voltage Gain ( $A_V$ ) = \_\_\_\_\_ (in mid band region).
3. Bandwidth (BW) = \_\_\_\_\_ Hz.
4. Figure of merit (FM) = \_\_\_\_\_ Hz.
5. Input impedance ( $Z_i$ ) = \_\_\_\_\_, Output Impedance ( $Z_o$ ) = \_\_\_\_\_.

$\Omega$

7 Block, Circuit, Model Diagram, Reaction Equation, Expected Graph



8 Observation Table, Look-up Table, Output

Frequency in Hz	$V_o$ in Volt	$A_V = \frac{V_o}{V_i}$	Gain in dB = $20 \log A_V$

9 Sample Calculations

**Design:**

Given  $V_{CE} = 5V$  and  $I_{C2} = 2mA$  Assume  $\beta = 100$   
 $V_{CC} = 2V_{CE} = 10V$ ,



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$$\text{Let } V_{RE} = 10\% V_{CC} = 1 \text{ V}$$

$$R_E = \frac{V_{RE}}{I_C + I_B}$$

$$I_B = \frac{I_C}{\beta} = \frac{2 \text{ mA}}{100} = 20 \mu \text{ A}$$

$$R_E = \frac{1}{2 \text{ mA} + 20 \mu \text{ A}} = 495 \Omega$$

$$\text{Choose } R_E = 470 \Omega$$

Apply KVL to Collector Loop

$$V_{CC} - I_C R_C - V_{CE} - V_E = 0$$

$$R_C = V_{CC} - \frac{V_{CE} - V_E}{I_C} = \frac{10 - 5 - 1}{2 \text{ m}} \text{ } \checkmark$$

$$R_C = 2 \text{ K } \Omega \quad \text{Choose } R_C = 1.8 \text{ K } \Omega$$

$$\text{Let } I_{R1} = 10 I_B = 10 \times 20 \mu \text{ A} = 200 \mu \text{ A}$$

$$V_{R2} = V_{BE} + V_E = 0.6 + 1 = 1.6 \text{ V} \quad (\text{Since transistor is silicon make } V_{BE} = 0.6 \text{ V})$$

$$R_2 = \frac{V_{R2}}{I_{R1} - I_B} = \frac{1.6 \text{ V}}{200 \mu + 20 \mu} = 7.2 \text{ K } \Omega$$

$$\text{Choose } R_2 = 8.2 \text{ K } \Omega$$

$$R_1 = \frac{V_{CC} - V_{R2}}{I_{R1}} = \frac{10 - 1.6}{200 \mu \text{ A}} = 42 \text{ K } \Omega$$

$$\text{Choose } R_1 = 47 \text{ K } \Omega$$

The condition is that  $X_{CE} \ll R_E$

$$\text{Let } X_{CE} = \frac{R_E}{10}$$

$$\frac{1}{2\pi f C_E} = \frac{470}{10} \quad \text{Let } f = 100 \text{ Hz}$$

$$C_E = 33 \mu \text{ F}$$

$$\text{Choose } C_E = 47 \mu \text{ F}$$

$$\text{Also } C_{C1} = C_{C2} = 0.47 \mu \text{ F}$$



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10	Graphs, Outputs	<p style="text-align: center;"><math>f_L = \text{Lower cutoff frequency}</math>    <math>f_H = \text{Higher cutoff frequency}</math></p>
11	Results Analysis	<p>&amp; Quiescent point: <math>V_{CE} = \dots</math>, <math>I_C = \dots</math>.</p> <ol style="list-style-type: none"> <li>Voltage Gain (<math>A_V</math>) = _____ (in mid band region).</li> <li>Bandwidth (BW) = _____ Hz.</li> <li>Figure of merit (FM) = _____ Hz.</li> <li>Input impedance (<math>Z_i</math>) = _____ <math>\Omega</math>, Output Impedance (<math>Z_o</math>) = _____ <math>\Omega</math>.</li> </ol>
12	Application Areas	Used in low noise amplifiers used in radio frequencies transceiver circuits
13	Remarks	
14	Faculty Signature with Date	

### Experiment 03 : COLPITTS OSCILLATOR AND CRYSTAL OSCILLATOR

-	<b>Experiment No.:</b>	3	<b>Marks</b>	10	<b>Date Planned</b>		<b>Date Conducted</b>	
1	Title	<b>COLPITTS OSCILLATOR AND CRYSTAL OSCILLATOR</b>						
2	Course Outcomes	Design analog circuits using BJT/FETs and evaluate their performance characteristics.						
3	Aim	To Design and set-up the following tuned oscillator circuits using BJT, and determine the frequency of oscillation.  (a) Colpitts Oscillator (b) Crystal Oscillator						
4	Material Equipment Required	/						
		<b>Sl. No.</b>	<b>Particulars</b>	<b>Range</b>				<b>Quantity</b>
		1.	Transistor (SL100/CL100)	-				1
		2.	Crystal	2MHz				1
		3.	Resistor	47K $\Omega$ , 1.8K $\Omega$ , 8.2K $\Omega$ , 470 $\Omega$				1,1,1,1
		4.	Capacitor	0.47 $\mu$ F, 47 $\mu$ F				2,1
		5.	Decade Capacitance Box	-				2

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		6.	CRO	-	1
		7.	Power Supply	0-30V	1
		8.	Multimeter	-	1
		9.	Spring Board	-	1
		10.	Connecting Wires	-	-
		<b>CRYSTAL OSCILLATOR</b>			
		<b>Sl. No.</b>	<b>Particulars</b>	<b>Range</b>	<b>Quantity</b>
		1.	Transistor (SL100/CL100)	-	1
		2.	Crystal	2MHz	1
		3.	Resistor	47KΩ, 1.8KΩ, 8.2KΩ, 470Ω	1, 1, 1, 1
		4.	Capacitor	0.47μF, 47 μF	2, 1
		5.	Decade Capacitance Box	-	2
		6.	CRO	-	1
		7.	Power Supply	0-30V	1
		8.	Multimeter	-	1
		9.	Spring Board	-	1
		10.	Connecting Wires	-	-
5	Theory, Formula, Principle, Concept	<p><b>Colpitts Oscillator:</b> A Colpitts oscillator is an electronic oscillator that uses a combination of inductor and capacitor to produce an output of certain frequency. It consists of a gain device with its output connected to the input in a feedback loop consisting of a parallel LC circuit.</p> <p>The Colpitts oscillator uses a capacitance voltage divider network as its feedback source. The two capacitors are connected across a single common inductor. The advantage of this type of capacitance circuit configuration is that with less and mutual inductance within the tank circuit, frequency stability of the oscillator is improved along with more simple design.</p> <p style="text-align: center;"><b>CRYSTAL OSCILLATOR:</b></p> <p>Crystal oscillators are used in order to get stable sinusoidal signals despite of variations in temperature, humidity, transistor and circuit parameters. A piezo electric crystal is used in this oscillator as resonant tank circuit. Crystal works under the principal of piezo-electric effect. i.e., when an AC signal applied across the crystal, it vibrates at the frequency of the applied voltage. Conversely if the crystal is forced to vibrate it will generate an AC signal. Commonly used crystals are Quartz, Rochelle salt etc.</p>			
6	Procedure, Program, Activity,	<p>1. Connections are made as shown in the diagram.</p>			



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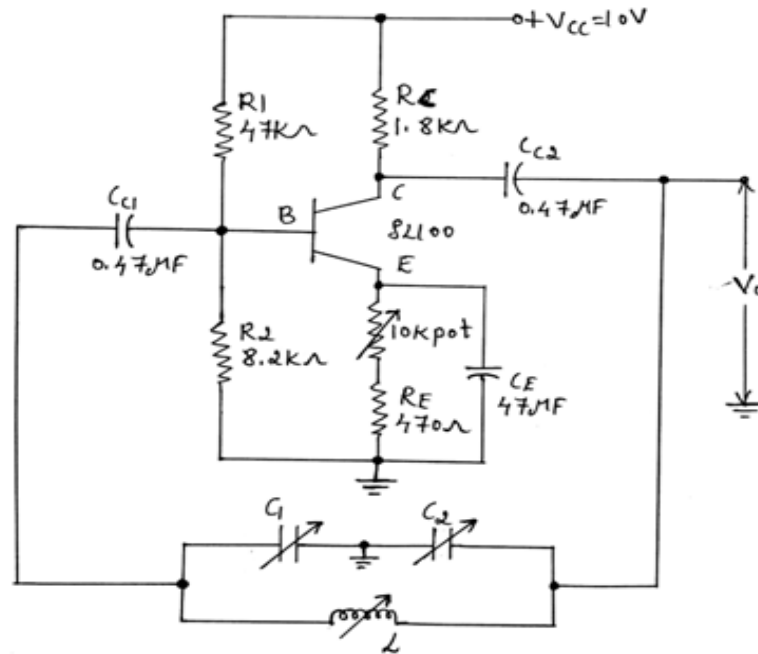
Algorithm,  
Pseudo Code

2. The quiescent point of the amplifier is verified for the designed value.
3. Observe the output waveform on CRO and measure the frequency.
4. Verify the output frequency with the theoretical frequency.

**Crystal oscillator**

5. Connections are made as shown in the diagram.
6. The quiescent point of the amplifier is verified for the designed value.
7. Observe the output wave form on CRO and measure the frequency.
8. Verify the frequency with the crystal frequency.

7 Block, Circuit,  
Model Diagram,  
Reaction  
Equation,  
Expected Graph

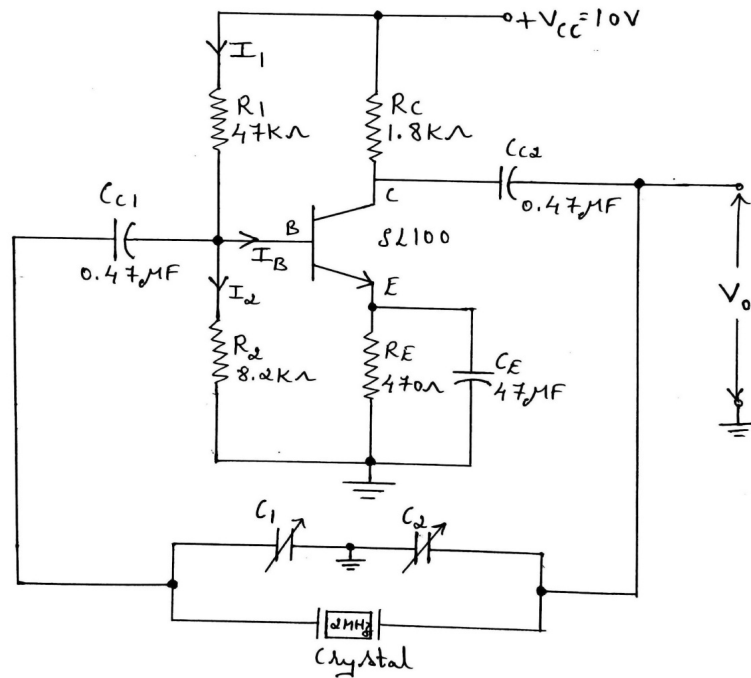


Crystal oscillator:



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8 Observation Table, Look-up Table, Output

9 Sample Calculations

**Colpitts Oscillator tank circuit design:**

Choose  $A=2$ ,  $f_o=100$  KHz

$$f_o = \frac{1}{2\pi\sqrt{LC_{eq}}} \text{ -----(c)}$$

$$C_{eq} = L_1 + L_2 \text{ -----(d)}$$

Condition for oscillation

$$A\beta \geq 1$$

$$A = \frac{C_1}{C_2} \Rightarrow C_1 = 2C_2$$

Find  $C_{eq}$  from (d) and L from (c)

crystal oscillator:

**Design:**

Given,  $V_{CE}=5V$  and  $I_C=2mA$ , Assume  $\beta = 100$

$$V_{CC} = 2V_{CE} = 2 \times 5 = 10V$$

Let  $V_{RE} = 10\%V_{CC} = 1V$



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$$R_E = \frac{V_{RE}}{I_C + I_B}$$
$$I_B = \frac{I_C}{\beta} = \frac{2 \text{ mA}}{100} = 20 \mu A$$
$$R_E = \frac{1}{2 \text{ mA} + 20 \mu} = 495 \Omega$$

Choose  $R_E = 470 \Omega$

Apply KVL to collector loop

$$V_{CC} - I_C R_C - V_{CE} - V_E = 0$$
$$R_C = \frac{V_{CC} - V_{CE} - V_E}{I_C} = \frac{10 - 5 - 1}{2 \text{ m}} = 2 \text{ K}\Omega$$

Choose  $R_C = 1.8 \text{ K}\Omega$

Let  $IR_1 = 10 I_B = 10 \times 20 \mu A = 200 \mu A$

$$VR_2 = V_{BE} + V_E = 0.6 + 1 = 1.6 \text{ V} \quad (\text{Since transistor is silicon make } V_{BE} = 0.6 \text{ V})$$
$$R_2 = \frac{VR_2}{IR_1 - I_B} = \frac{1.6}{200 \mu A + 20 \mu A}$$
$$R_2 = 7.2 \text{ K}\Omega \quad \text{Choose } R_2 = 8.2 \text{ K}\Omega$$
$$R_1 = \frac{(V_{CC} - VR_2)}{IR_1} = \frac{(10 - 1.6)}{200 \mu A} \quad R_1 = 42 \text{ K}\Omega$$

Choose  $R_1 = 47 \text{ K}\Omega$

$$X_{CE} \ll R_E$$
$$X_{CE} = \frac{R_E}{10}$$
$$\frac{1}{2\pi f C_E} = \frac{470}{10}$$

Let  $f = 100 \text{ Hz}$

$$C_E = 33 \mu F \quad \text{Choose } C_E = 47 \mu F$$

Choose  $CC_1 = CC_2 = 0.47 \mu F$



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10	Graphs, Outputs	
11	Results Analysis	& Colpitts Oscillator: Theoretical frequency (f) = _____ Hz Practical frequency (f) = _____ Hz.  For Crystal = _____ Hz. For Practical = _____ Hz.
12	Application Areas	
13	Remarks	
14	Faculty Signature with Date	

.....

### Experiment 04 : SECOND ORDER BUTTERWORTH LOW PASS AND HIGH PASS FILTER

-	Experiment No.:	4	Marks	10	Date Planned	Date Conducted
1	Title	SECOND ORDER BUTTERWORTH LOW PASS AND HIGH PASS FILTER				
2	Course Outcomes	Design analog circuits using OPAMPs and 555 timer for different applications				
3	Aim	To design a second order butter worth second order low pass and high pass filter for a given cut off frequency and draw the frequency response.				
4	Material Equipment Required	<b>Components Required:</b> <ol style="list-style-type: none"> <li>1. IC 741 Op-Amp</li> <li>2. Resistors – As per the design</li> <li>3. Capacitor – As per the design</li> <li>4. Power supply</li> <li>5. Signal generator</li> <li>6. CRO</li> </ol>				
5	Theory, Formula, Principle, Concept	<b>Low pass filter:</b>				

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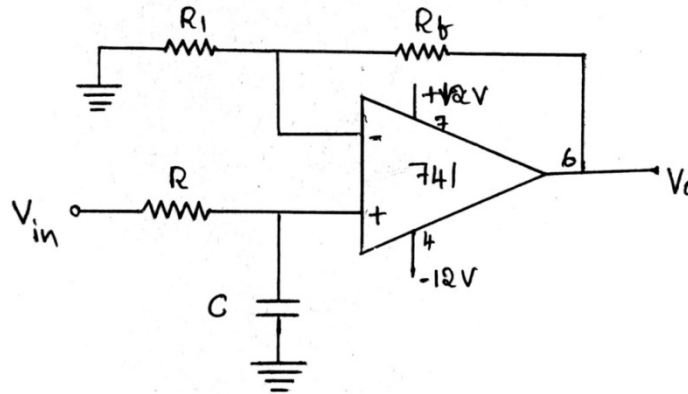


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Low pass filter allows only low frequency signal to pass through them. A low pass filter can be a combination of capacitance, inductance or resistance to produce high attenuation above a specified frequency & little or no attenuation below that frequency. The frequency at which the transition occurs is called cut-off frequency.

- A first order low pass Butterworth filter uses RC network for filtering. The op-Amp is used in non-inverting configuration.



- The first order filter can be connected to second order LPF by using additional RC network as shown in fig1.

**2<sup>nd</sup> order:**

- The stop-band response in 2<sup>nd</sup> order LPF is 40dB/decade. At low frequency, both capacitors appear open and the circuit becomes a non-

inverting amplifier  $\left( \dots X_C = \frac{1}{2\pi FC} \right)$

- As frequency increase, the gain eventually starts to decrease until it is down 3dB at the cutoff frequency. As frequency increase the cut off frequency, the o/p is attenuated.

**b) High pass filter:**

High pass filter passes high frequency signals to pass through it. Again frequency sensitive components such as capacitors & inductors are used in conjunction with the resistors.

The first order high pass filter is formed from first order low pass filter by interchanging the R&C components & second order HPF filter can be obtained from 2<sup>nd</sup> order LPF by interchanging R&C as shown in Fig3.

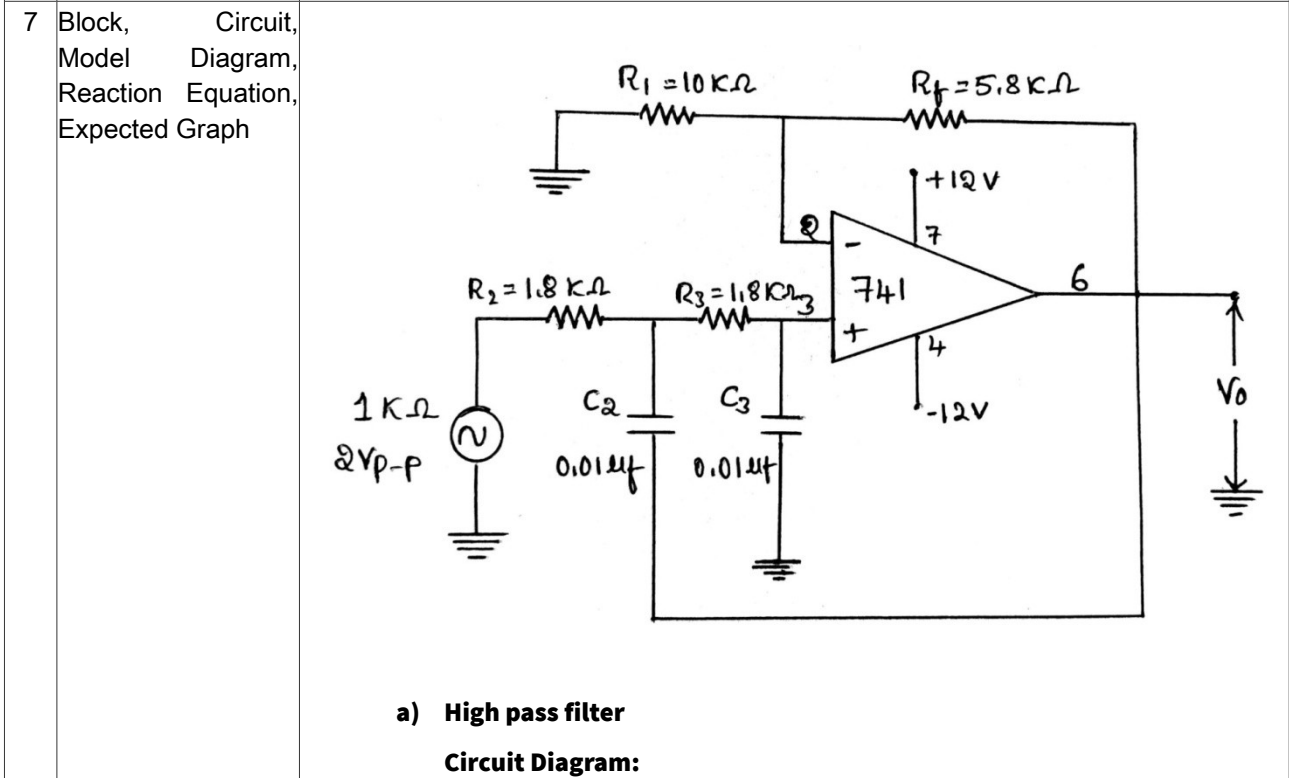


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Here as frequency increase, the capacitors act as short & the circuit behaves as amplifier with gain  $A_f$  but when the frequency is low, the capacitor act as open circuit & hence very little signal passes through the circuit.

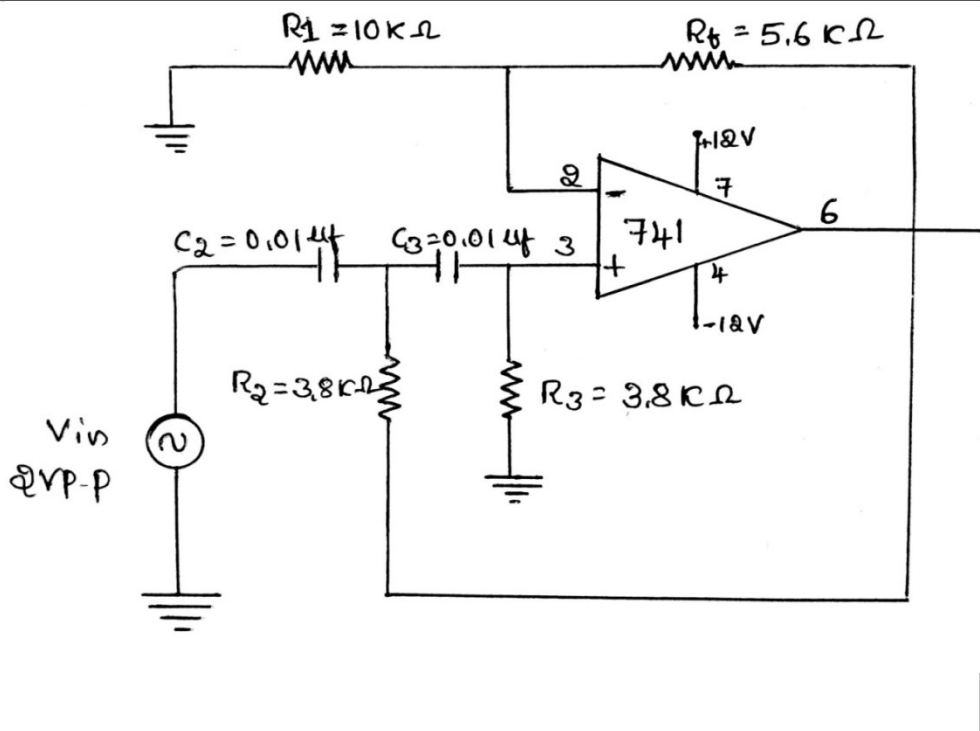
- |   |  |  |
|---|--|--|
| 6 | Procedure, Program, Activity, Algorithm, Pseudo Code | <ol style="list-style-type: none"> <li>1. Rig up the circuits as show in the diagrams</li> <li>2. apply the input as specified</li> <li>3. vary the input signal frequency and note down the output voltages and calculate the gain convert the gain in db</li> <li>4. plot the graph frequency/ gain in db and observe the cut-off points and calculate the roll off</li> </ol> |
|---|--|--|





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8 Observation Table, Look-up Table, Output

$$V_i = 2V_{p-p}$$

Frequency (Hz)	$V_o$ (V)	$A_v = \frac{V_o}{V_i}$	$20 \text{ LOG } A_v$

9 Sample Calculations

**LPF:**  
 The higher cutoff frequency that is the frequency at which signal falls (3dB) below is given by

$$\text{for LPH: } f_H = \frac{1}{2\pi\sqrt{R_2 R_3 C_2 C_3}}$$

**HPF:**  
 The lower cutoff frequency that is the frequency at which signal strength falls 3dB below this is given by





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$$f_L = \frac{1}{2\pi\sqrt{R_2 R_3 C_2 C_3}}$$

**b) Low pass filter**

Design 2<sup>nd</sup> order LPF to obtain  $f_H = 9\text{KHz}$

Solution:

$$f_H = \frac{1}{2\pi\sqrt{R_2 R_3 C_2 C_3}}$$

Let  $R_2 = R_3 = R$  &  $C_2 = C_3 = C$

$$\therefore f_H = \frac{1}{2\pi RC}$$

Choose  $C = 0.01\mu\text{f}$

$$\therefore R = \frac{1}{2\pi f_H C} = \frac{1}{2\pi * 9\text{K} * 0.01\mu\text{f}} = 1.768\text{k}\Omega$$

$\therefore$  Choose  $R = 1.8\text{k}\Omega$

The pass band gain of 2<sup>nd</sup> order filter = 1.586

$$A_f = 1 + \frac{R_f}{R_1} = 1.586$$

$$\therefore \frac{R_f}{R_1} = 0.586$$

Let  $R_1 = 10\text{k}\Omega$  & hence  $R_f = 5.86\text{k}\Omega$

$$R_1 \implies R_f = 5.86\text{k}\Omega$$

**Design:**

Design HPF to obtain  $f_H = 4\text{KHz}$

Solution:

$$f_H = \frac{1}{2\pi\sqrt{R_2 R_3 C_2 C_3}}$$

Let  $R_2 = R_3 = R$  &  $C_2 = C_3 = C$  & hence

$$f_H = \frac{1}{2\pi RC}$$

Let  $C = 0.01\mu\text{f}$

$$\therefore R = \frac{1}{2\pi f_H C} = \frac{1}{2\pi * 4\text{K} * 0.01\mu\text{f}} = 3.9\text{k}\Omega$$

Hence choose  $R = 3.8\text{k}\Omega$



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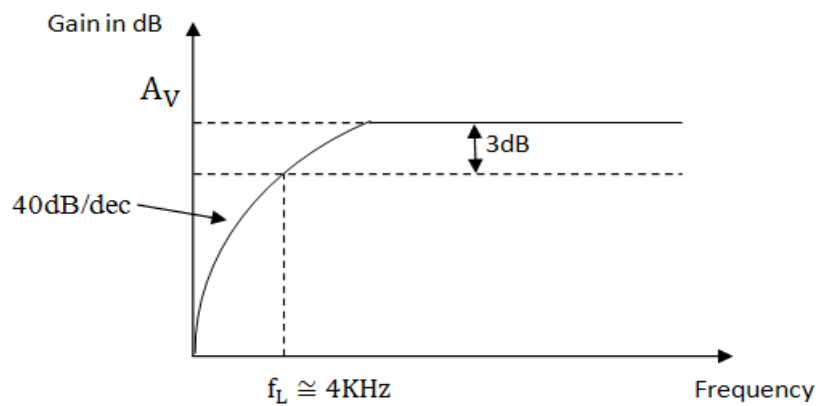
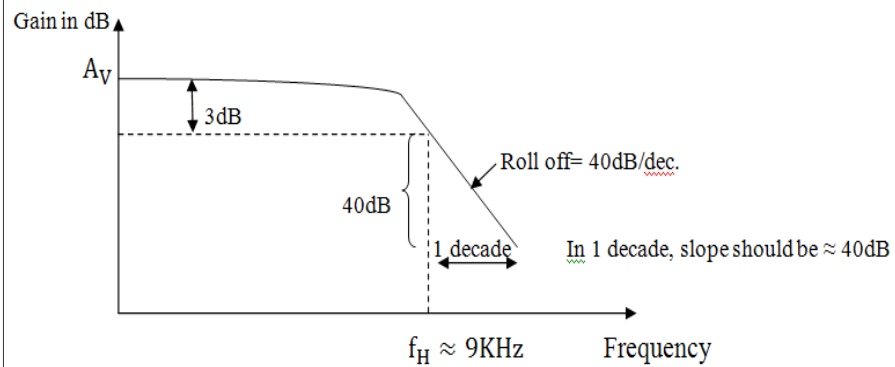
The pass band gain of 2<sup>nd</sup> order filter is 1.586

$$\therefore A_f = 1 + \frac{R_f}{R_1} = 1.586$$

$$\frac{R_f}{R_1} = 0.586$$

Let  $R_1 = 10K\Omega$  & hence  $R_f = 0.586 R_1 \implies R_f = 5.8K\Omega$

10 Graphs, Outputs



11 Results & Analysis

**Result:**

**LPF:** 1) Cut off frequency for Low pass filter:

Theoretical	Practical
9 KHz	

2) Roll off rate:

Theoretical	Practical
40dB/dec	



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		<p><b>HPF:</b> 1) Cut off frequency for HPF:</p> <table border="1"> <tr> <th>Theoretical</th> <th>Practical</th> </tr> <tr> <td>4 KHz</td> <td></td> </tr> </table> <p>2) Roll off rate:</p> <table border="1"> <tr> <th>Theoretical</th> <th>Practical</th> </tr> <tr> <td>40dB/dec</td> <td></td> </tr> </table>	Theoretical	Practical	4 KHz		Theoretical	Practical	40dB/dec	
Theoretical	Practical									
4 KHz										
Theoretical	Practical									
40dB/dec										
12	Application Areas	Used in generation of high frequencies sinusoidal signal used in computer , instrumentation and in digital systems								
13	Remarks									
14	Faculty Signature with Date									

.....

### Experiment 05 : ADDER, INTEGRATOR AND DIFFERENTIATOR USING OP-AMP

-	Experiment No.:	5	Marks	10	Date Planned		Date Conducted	
1	Title	<b>ADDER, INTEGRATOR AND DIFFERENTIATOR USING OP-AMP</b>						
2	Course Outcomes	Design analog circuits using OPAMPs and 555 timer for different applications						
3	Aim	To design adder, integrator and differentiator circuit for given specification using Op-Amp.						
4	Material Equipment Required	<ol style="list-style-type: none"> <li>IC 741 -Op-Amp</li> <li>Resistors – as per the design</li> <li>Capacitor – as per the design</li> <li>Dc power supply</li> <li>Signal generator</li> <li>CRO</li> <li>Digital multimeter/Voltmeter</li> </ol>						
5	Theory, Formula, Principle, Concept	<p><b>Adder:</b></p> <p>The most common application of Op-Amp is the summing-amplifier (or adder) circuit. Fig. 1 shows the inverting configuration of summing circuit with 2 inputs <math>V_1</math> and <math>V_2</math>. Depending on the relationship between <math>R_f</math>, the feedback resistor and the input resistor <math>R_1</math> and <math>R_2</math>, the circuit can be used as summing</p>						

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amplifier, scaling amplifier or averaging Amplifier. The input expression for the circuit can be written as (or obtained as)

$$I_F = I_1 + I_2$$

$$\frac{-V_o}{R_f} = \frac{V_1}{R_1} + \frac{V_2}{R_2}$$

$$\therefore V_o = - \left[ \frac{R_f}{R_1} V_1 + \frac{R_f}{R_2} V_2 \right] \dots\dots\dots (1)$$

- If  $R_1 = R_2 = R_f = R$  in (1), then  
 $V_o = - (V_1 + V_2)$  ---- summing amplifier. Here , the o/p voltage is equal to negative sum of all the inputs. Hence circuit act as summing amplifier.
- If  $R_1, R_2, R_f$  different , then the circuit is called scaling amplifier.
- If  $R_1 = R_2 = R$  & if  $\frac{R_f}{R} = \frac{1}{2}$  , then the circuit can be used as an averaging circuit.

**2. Integrator:**

A circuit in which the output voltage is the integral of the input voltage is called integrator as shown in Fig2.

Relationship between voltage and current through capacitor is given by

$$i_c = C \frac{dV_c}{dt}$$

Applying Kirchhoff's law,

$$i_1 \simeq i_f$$

$$\therefore \frac{V_{in}}{R_1} = C_F \frac{d(-V_o)}{dt}$$

$$\therefore V_o = - \frac{1}{R_1 C_F} \int_0^t V_{in} dt + const$$

- As seen ,the o/p voltage is directly proportional to the negative of the input voltage and inversely proportional to time constant  $R_1 C_F$ .
- If  $V_{in} = 0$ , then input offset voltage and the capacitor  $C_T$  produce error



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		<p>voltage at output of integrator. To reduce the error voltages, resistor <math>R_F</math> is connected across the feedback capacitor <math>C_F</math>.</p> <ul style="list-style-type: none"> <li>• Addition of <math>R_F</math> improves stability and low frequency roll off problems and hence minimizes the variations in the output voltage.</li> <li>• The input signal will be integrated properly if the time period T of the signal is larger than or equal to <math>R_F C_F</math>.</li> <li>• <math>\therefore T \geq R_F C_F</math>.</li> </ul> <p><b>3. Differentiator:</b></p> <p>Differentiator performs the mathematical operation of differentiation. The output voltage can be expressed in Fig. 3 as</p> $i_c = i_F$ $\therefore C_1 \cdot \frac{d(V_{in})}{dt} = - \frac{V_o}{R_F}$ $\therefore V_o = - R_F C_1 \frac{d(V_{in})}{dt}$ <ul style="list-style-type: none"> <li>• The gain <math>\frac{R_F}{X_{C1}}</math> increase with increase in frequency. Also, the input impedance <math>X_{C1}</math> decreases with increase in frequency which makes circuit susceptible to high frequency noise.</li> <li>• The stability &amp; high frequency noise problem can be corrected by two components <math>C_F</math> &amp; <math>R_1</math>.</li> <li>• The input signal will be differentiated properly if the time period T of the input signal is larger than or equal to <math>R_F C_1</math>.</li> <li>• <math>\therefore T \geq R_F C_1</math>.</li> </ul>
6	<p>Procedure, Program, Activity, Algorithm, Pseudo Code</p>	<p><b>Adder :</b></p> <ol style="list-style-type: none"> <li>1. Before wiring the circuit, check the components for its working .</li> <li>2. Connect the circuit as shown in Fig.1.</li> <li>3. Set the input voltages <math>V_1</math> and <math>V_2</math> and measure the output voltage <math>V_o</math> using multimeter.</li> <li>4. Compare theoretical and practical output voltages</li> </ol>



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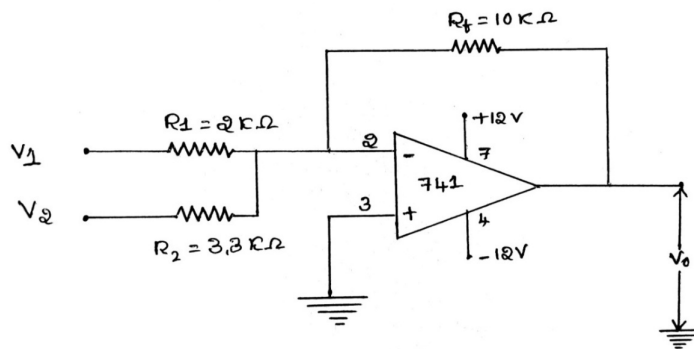
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**5. Integrator Procedure:**

6. Check all the components for its working.
  7. Make the connections as shown in Fig.2.
  8. Set the input voltage using signal generator to 4V (or 2V) peak to peak square wave at 1KHz frequency.
  9. Observe the input & output signals of circuit on CRO.
  10. Sketch the output for  $RC = 10T$  &  $RC = T$  &  $RC = 0.1T$ .
- Differentiator”
11. Check all the components for its working.
  12. Make the circuit diagram as show in Fig.3.
  13. Set the input voltage using signal generator to 4V (or 2V) peak to peak triangular wave at 1KHz frequency.
  14. Sketch the output for  $RC = 10T$ ,  $0.025T$  & varying R.
  15. Observe the output signal on CRO

7 Block, Circuit, Model Diagram, Reaction Equation, Expected Graph

Adder circuit

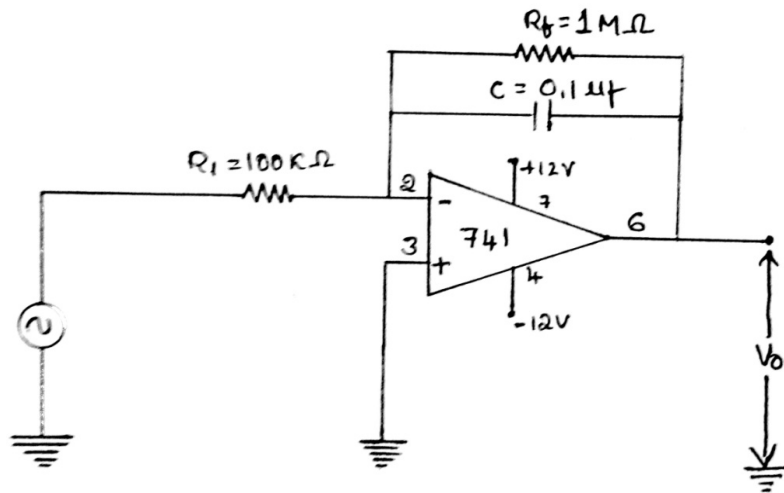


Integrator:

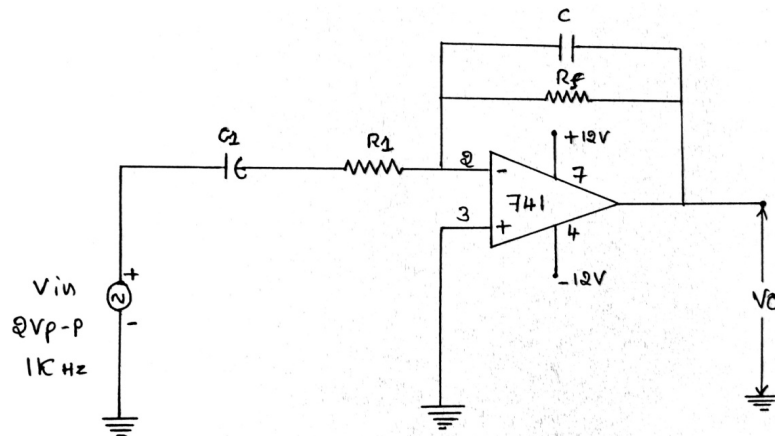


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Differentiator:



8 Observation Table, Look-up Table, Output

Input voltage (V)		Output voltage (Vo)	
V <sub>1</sub>	V <sub>2</sub>	Theoretical V <sub>O</sub>	Practical V <sub>O</sub>

9 Sample Calculations

**ADDER Design:**

Design a circuit to obtain the output voltage

$$V_o = -(5V_1 + 3V_2)$$

**Solution:**

Given  $V_o = -(5V_1 + 3V_2)$

But  $V_o = -\left[\frac{R_f}{R_1} V_1 + \frac{R_f}{R_2} V_2\right]$



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$$\text{Given } \frac{R_f}{R_1} = 5 \text{ \& } \frac{R_f}{R_2} = 3$$

$$\therefore R_f = 5R_1, R_f = 3R_2$$

Choose  $R_f = 10 \text{ K}\Omega$

$$\therefore R_1 = \frac{R_f}{5} = \frac{10k}{5} = 2\text{K}\Omega \dots\dots\dots R_1 = 2\text{K}\Omega$$

$$R_2 = \frac{R_f}{3} = \frac{10k}{3} = 3.3\text{K}\Omega \dots\dots\dots R_2 = 3.3\text{K}\Omega$$

**Integrator Design:**

Given T = 1ms

a) Let RC = 10T

Choose C = 0.1µf

$$\therefore R = \frac{10T}{C} = \frac{10 * 1 * 10^{-3}}{0.1 * 10^{-6}}$$

$$R = 100\text{k}\Omega$$

b) If RC = T

$$\therefore R = \frac{T}{C} = \frac{1 * 10^{-3}}{0.1 * 10^{-6}}$$

$$\therefore R = 10\text{K}\Omega$$

c) If RC = 0.1 T

$$R = \frac{0.1 * 1 * 10^{-3}}{0.1 * 10^{-6}}$$

$$\mathbf{R = 1K}\Omega$$

**Differentor Design:**

Given T = 1ms

a) Let RC = 10T

Choose C = 0.1µf

$$\therefore R = \frac{10T}{C} = \frac{10 * 1 * 10^{-3}}{0.1 * 10^{-6}}$$

$$R = 100\text{k}\Omega$$

b) If RC = 0.025T

$$R = \frac{0.025T}{C} = \frac{0.025 * 1 * 10^{-3}}{0.1 * 10^{-6}}$$

$$R = 250\Omega$$





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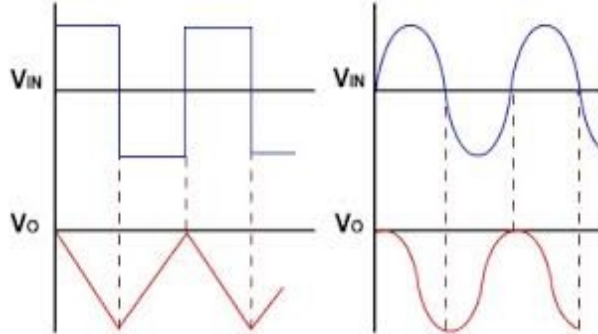
c) If  $RC = T$

$$R = \frac{T}{C} = \frac{1 \times 10^{-3}}{0.1 \times 10^{-6}}$$

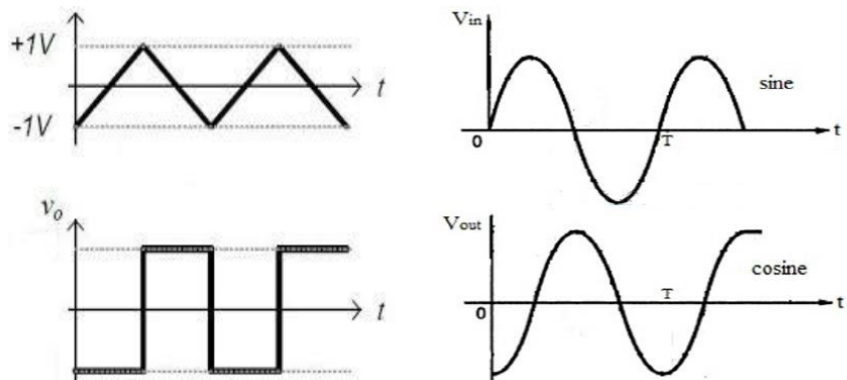
$$R = 10K\Omega$$

10 Graphs, Outputs

Integrator :



**Differentiator :**



11 Results & Analysis

1.Adder:

The obtained output voltage for inputs  $V_1$  and  $V_2$  are  
 $V_o = \dots\dots$  when  $V_1 = \dots\dots$  &  $V_2 = \dots\dots$



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		<p>2. The operation of integrator circuit is verified.</p> <p>3. Operation of differentiator circuit is verified.</p>
12	Application Areas	<p>The adder circuit is commonly used in</p> <ol style="list-style-type: none"> <li>1). Analog computers</li> <li>2). Audio mixers in which no of inputs are added or mixed to produce desired output.</li> </ol> <p>Integrator is commonly used in</p> <ol style="list-style-type: none"> <li>1. Analog computers.</li> <li>2. Analog to digital converter.</li> <li>3. Signal wave shaping circuits</li> </ol> <p>Differentiator is commonly used in</p> <ol style="list-style-type: none"> <li>1. Wave shaping circuits to detect high frequency component in input signal.</li> <li>2. Rate of change detector in FM modulators.</li> </ol>
13	Remarks	
14	Faculty Signature with Date	

### Experiment 06 : SCHMITT TRIGGER

-	Experiment No.:	6	Marks	10	Date Planned	Date Conducted
1	Title	<b>SCHMITT TRIGGER</b>				
2	Course Outcomes	Design analog circuits using OPAMPs and 555 timer for different applications				
3	Aim	Design and Testing of Schmitt trigger circuit for different hysteresis value				
4	Material Equipment Required	IC trainer, signal generator, CRO. Resisters, OP AMP, Patch Chords, Digital multimeter.				
5	Theory, Formula, Principle, Concept	<p>A Schmitt trigger is a comparator circuit with hysteresis implemented by applying positive feedback to the non inverting input of a comparator or differential amplifier. It is an active circuit which converts an analog input signal to a digital output signal. In the non-inverting configuration, when the input is higher than a chosen threshold, the output is high. When the input is below a different (lower) chosen threshold the output is low, and when the input is between the two levels the output retains its value. This dual threshold action is called hysteresis and implies that the Schmitt trigger possesses memory and can act as a bistable multivibrator.</p> <p>Schmitt trigger devices are typically used in signal conditioning applications to remove noise from signals used in digital circuits, particularly mechanical contact bounce. They are also used in closed loop negative feedback configurations to</p>				

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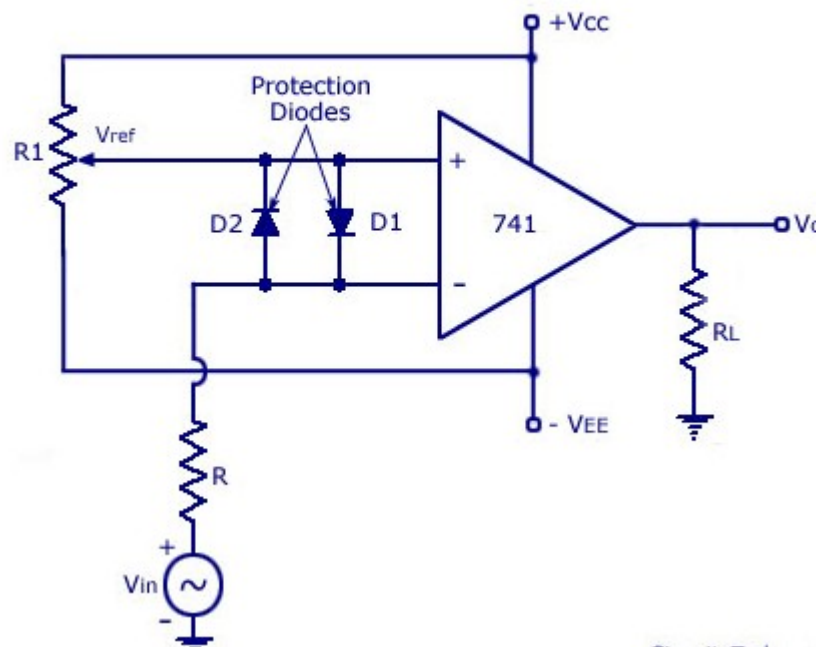
implement relaxation oscillators, used in function generators and switching power supplies.

### Test a Comparator circuit

An inverting 741 IC op-amp comparator circuit is shown in the figure below. It is called an inverting comparator circuit as the sinusoidal input signal  $V_{in}$  is applied to the inverting terminal. The fixed reference voltage  $V_{ref}$  is give to the non-inverting terminal (+) of the op-amp. A potentiometer is used as a voltage divider circuit to obtain the reference voltage in the non-inverting input terminal. Bothe ends of the POT are connected to the dc supply voltage +VCC and -VEE. The wiper is connected to the non-inverting input terminal. When the wiper is rotated to a value near +VCC,  $V_{ref}$  becomes more positive, and when the wiper is rotated towards -VEE, the value of  $V_{ref}$  becomes more negative. The waveforms are shown below.

### Inverting Comparator circuit

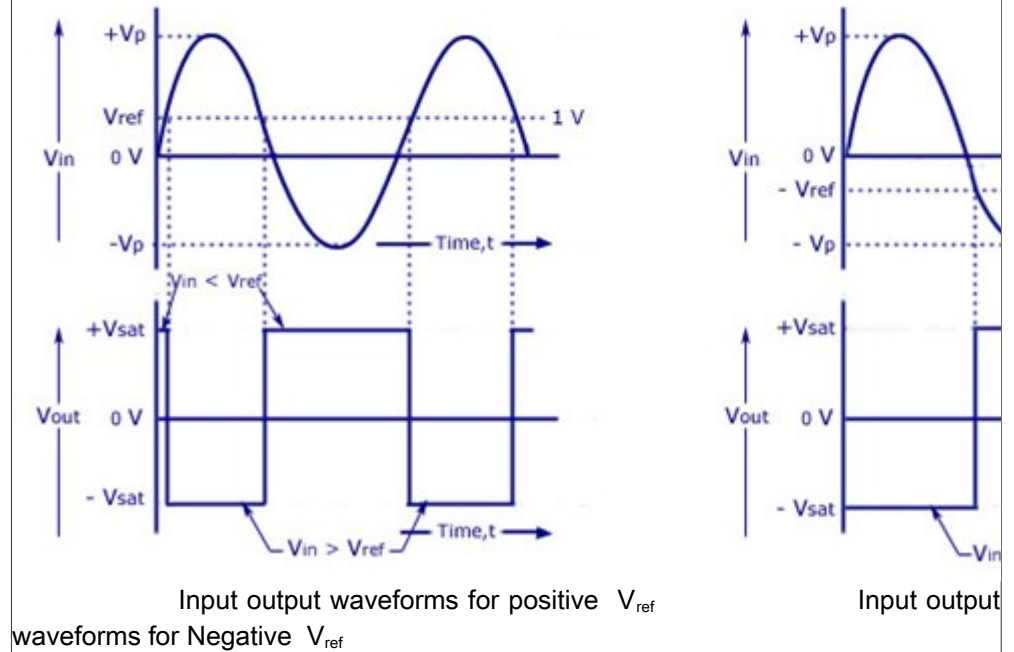
Inverting Comparator Circuit





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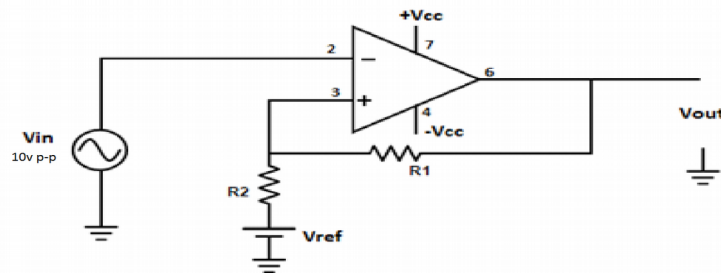
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Procedure, Program, Activity, Algorithm, Pseudo Code

- Rig up the circuit as shown in figure 1.
1. Apply an input of 10V p-p, 1 kHz to the input terminals.
  2. Observe the output waveform and measure the practical values of  $V_{UTP}$ ,  $V_{LTP}$  in the x-y mode or the y-t mode. Also measure the value of  $V_{SAT}$ .

7 Block, Model, Circuit, Diagram, Reaction Equation, Expected Graph



8 Observation Table, Look-up Table, Output

9 Sample Calculations

**Design: Formulae:**

$$V_{UTP} = V_{SAT} \cdot \frac{R_2}{(R_1+R_2)} + V_{REF} \cdot \frac{R_1}{(R_1+R_2)} \dots\dots\dots (1)$$

$$V_{LTP} = V_{SAT} \cdot \frac{-R_2}{(R_1+R_2)} + V_{REF} \cdot \frac{(R_1)}{(R_1+R_2)} \dots\dots\dots (2)$$

Adding (1) and (2)

$$V_{UTP} + V_{LTP} = V_{REF} \frac{2R_1}{(R_1+R_2)} \dots\dots\dots (3)$$

$$V_{UTP} - V_{LTP} = V_{SAT} = \frac{2R_2}{(R_1+R_2)} \dots\dots\dots (4)$$



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	$V_{UTP} = 4V, V_{LTP} = 2V, V_{SAT} = 12V$ From (4) $R_2 / (R_1 + R_2) = 0.08333$ Therefore $R_1 / R_2 = 11$ Choose $R_2 = 1 \text{ k ohm}, R_1 = 11 \text{ k ohm}$ From (3) $V_{REF} = (V_{UTP} + V_{LTP}) \cdot (R_1 + R_2) / (2 R_1), V_{REF} = (6 \cdot 12 \times 10^3) / (2 \cdot 11 \times 10^3)$ $V_{REF} = 3.272V$
10	Graphs, Outputs 
11	Results & Analysis $V_{UTP} \text{ (Practical)} = \dots\dots\dots \text{Volts.}$ $V_{LTP} \text{ (Practical)} = \dots\dots\dots \text{Volts.}$ $V_{+sat} = \dots\dots\dots \text{Volts.}$ $V_{-sat} = \dots\dots\dots \text{Volts.}$
12	Application Areas Used in signal conditioning to remove noise
13	Remarks
14	Faculty Signature with Date

### Experiment 07 : R-2R DAC USING OP-AMP

-	Experiment No.:	8	Marks	10	Date Planned	Date Conducted
1	Title	<b>R-2R DAC USING OP-AMP</b>				
2	Course Outcomes	Design analog circuits using OPAMPs and 555 timer for different applications				
3	Aim	.1. To design 4 bit R-2R digital to analog convertor using Op-Amp and verify its operation using input toggle switch. 2. To generate staircase waveform using mod-16 counter.				
4	Material Equipment Required	<b>Components Required:</b> 1. IC 741 Op-Amp 2. Resistors –As per the design 3. Dc power supply 4. IC trainer kit				



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		<p>5. CRO</p> <p>6. Millimeter / Voltmeter</p> <p>7. IC7493 or equivalent</p>
5	Theory, Formula, Principle, Concept	<p>R-2R DAC is shown in Fig. 1. It consists of only two resistors R and 2R forming ladder network and an Op-Amp acting as voltage follower. Here <math>D_0, D_1, D_2</math> &amp; <math>D_3</math> are digital inputs which are controlled by the Switches <math>S_0, S_1, S_2,</math> &amp; <math>S_3</math>. When the digital input is '1', then the corresponding switch connects the resistor 2R to <math>V_{ref}</math> and when digital input is '0', then the switch connects the resistor 2R to the ground line. Since the ladder is composed of linear resistors, it is a linear network and hence principle of superposition can be used to obtain the output voltage.</p> <ul style="list-style-type: none"> <li>The analog output voltage <math>V_o</math> for 4 bit DAC can be written as</li> </ul> $V_o = [2^3 D_3 + 2^2 D_2 + 2^1 D_1 + 2^0 D_0] V_1$ <p>where</p> $V_1 = \left[ \frac{V_{ref}}{2^N} \right] * \left[ \frac{2R}{R+2R} \right] = \frac{V_{ref}}{2^N} \left[ \frac{2R}{3R} \right] = \frac{V_{ref}}{2^N} \left[ \frac{2}{3} \right]$ <p>Since N=4 [4 bit DAC], <math>V_1</math> becomes</p> $V_1 = \frac{V_{ref}}{2^4} * \frac{2}{3}$ $V_1 = \frac{V_{ref}}{24}$ <p>∴ <math>V_o = [8 D_3 + 4 D_2 + 2 D_1 + D_0] * \frac{V_{ref}}{24}</math></p>
6	Procedure, Program, Algorithm, Code, Activity, Pseudo Code	<ol style="list-style-type: none"> <li>Verify the components for its working.</li> <li>Make the connection as shown in Fig.1.</li> <li>For different digital inputs, measure the output voltage using multimeter.</li> <li>Verify whether the theoretical values is matching with practical values &amp; plot the graph of input V/s output.</li> </ol> <p><b>Procedure:</b></p> <ol style="list-style-type: none"> <li>Check the components for its working.</li> <li>Make connection as shown in Fig. 3.</li> <li>Construct modulo16 counter using suitable IC like 7493 or 74193.</li> </ol>



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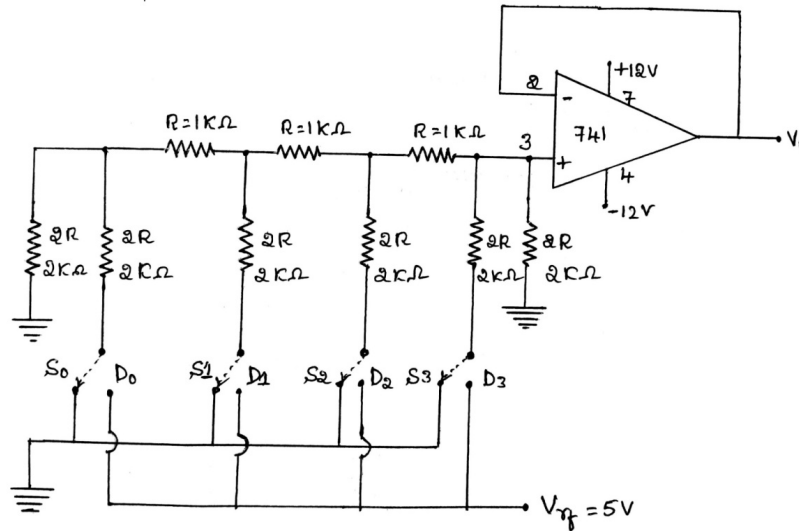
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- Apply clock (Say 1KHz or 10KHz ) and observe staircase waveform on CRO.
- Find resolution and sketch input and output waveform on graph sheet.

7 Block, Circuit, Model Diagram, Reaction Equation, Expected Graph

**R-2R ladder Network:**

**a). circuit diagram:**



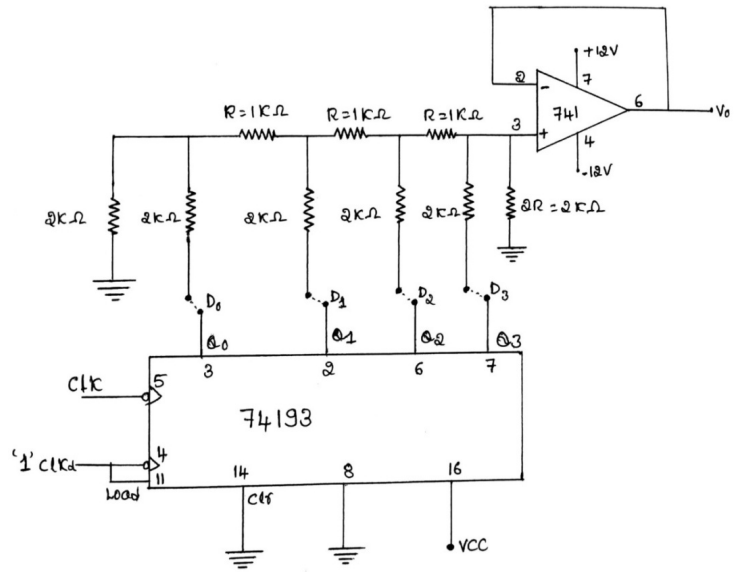
- To generate staircase waveform using DAC circuit & mod-16 counter

**Circuit diagram:**



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8 Observation Table, Look-up Table, Output

**Tabular column:**

Digital inputs				Decimal equivalent	Analog output voltage
$D_3$	$D_2$	$D_1$	$D_0$		Theoretical $V_0$ (V)
0	0	0	0	0	0
0	0	0	1	1	0.20833
0	0	1	0	2	0.4166
0	0	1	1	3	0.625
0	1	0	0	4	0.833
0	1	0	1	5	1.00416
0	1	1	0	6	1.25
0	1	1	1	7	1.45
1	0	0	0	8	1.66
1	0	0	1	9	1.875
1	0	1	0	10	2.05
1	0	1	1	11	2.29
1	1	0	0	12	2.50
1	1	0	1	13	2.708
1	1	1	0	14	2.916
1	1	1	1	15	3.125

**Tabular Column:**





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		<table border="1"> <tr> <th colspan="2">Step width</th> <th colspan="2">Resolution</th> </tr> <tr> <td>Ideal</td> <td>Obtained</td> <td>Ideal</td> <td>Obtained</td> </tr> <tr> <td></td> <td></td> <td></td> <td></td> </tr> </table> $\% \text{ Resolution} = \frac{\text{Stepsize}}{\text{Full scale}} * 100$ $= \frac{\text{Stepwidth}}{4} * 100$	Step width		Resolution		Ideal	Obtained	Ideal	Obtained				
Step width		Resolution												
Ideal	Obtained	Ideal	Obtained											
9	Sample Calculations	<p><b>Given: No of steps =15</b></p> <p><b>Solution:</b></p> <p>w.k.t, No of steps = <math>2^N - 1</math></p> $15 = 2^N - 1$ $14 = 2^N$ <p>Apply <math>\log_2</math> on both sides</p> $\log_2 14 = \log_2 2^N$ $\log_2 14 = N$ $\frac{\log 14}{\log 2} = N$ $N = 3.8$ <p><math>\therefore</math> Resolution = N</p> <p>Use 4 bit DAC with minimum step size as 0.208V.</p> <p>This is a 4 bit R-2R ladder N/W shown in Fig3.</p> <p>a) Given :N=4 and step size =0.5V</p> <p><b>Solution:</b></p> $\text{Minimum step size} = \frac{V_{\text{ref}}}{2^N} = \frac{5}{2^4} = 0.208V$ <p>Given step = 0.5V</p> $\therefore \text{Maximum o/p voltage} = V_{\text{omax}} = \text{step size} * \text{No of steps}$ $= 0.5 * (2^4 - 1)$ $= 0.5 * 15$ $= 7.5V$ <p>Given steps =0.5</p>												



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		<p> <math>\therefore \text{Gain} = A_f = \frac{\text{Given step}}{\text{minimum step}} = \frac{0.5}{0.208} = 2.4V</math> </p> <p>             But <math>A_f = 1 + \frac{R_f}{R_1}</math> </p> <p> <math>2.4 = 1 + \frac{R_f}{R_1}</math> </p> <p> <math>\therefore \frac{R_f}{R_1} = 1.4</math> </p> <p> <math>\therefore R_f = 1.4 R_1</math> </p> <p>             Let <math>R_1 = 10K\Omega</math> </p> <p> <math>\therefore R_f = 1.4 * 10K\Omega</math> </p> <p> <math>= 14K\Omega</math> </p> <p> <math>R_f \approx 15K\Omega</math> </p>
10	Graphs, Outputs	<p><b>Specimen graph:</b></p>
11	Results & Analysis	<p>1. The obtained resolution of DAC is .....</p> <p>2. Working of R-2R DAC is verified.</p>
12	Application Areas	<p>Typical applications for D/A converter include microcomputer interfacing, CRT graphics generations, programmable power supplies, digitally controlled gain circuits, digital filters, etc.,.</p>
13	Remarks	
14	Faculty Signature with Date	



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## Experiment 08 : ASTABLE AND MONOSTABLE MULTIVIBRATOR USING IC 555

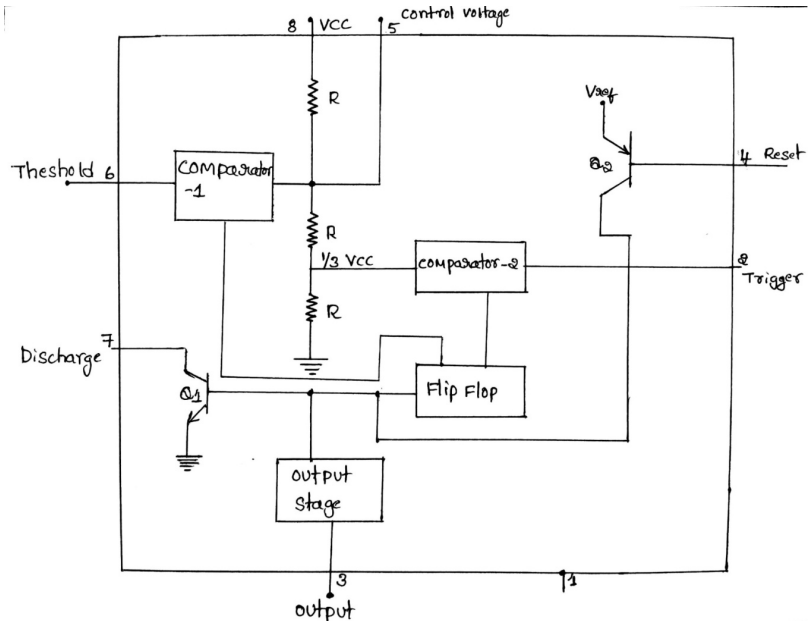
-	Experiment No.:	8	Marks	10	Date Planned	Date Conducted
1	Title	<b>ASTABLE AND MONOSTABLE MULTIVIBRATOR USING IC 555</b>				
2	Course Outcomes	Design analog circuits using OPAMPs and 555 timer for different applications				
3	Aim	1. Design Astable multi-vibrator using IC 555 timer to generate a clock frequency of 1KHz with 0.75 duty cycle (unsymmetrical) and 0.5 duty cycle (symmetrical). 2. Design monostable multi-vibrator using IC 555 timer.				
4	Material Equipment Required	1. 555 timer 2. Resistors – As per the design 3. Capacitors – As per the design 4. Power supply 5. Diode – 1N4001 6. Signal generator				
5	Theory, Formula, Principle, Concept	<p><b>Astable multivibrator:</b></p> <p>An astable multi-vibrator, often called a free running multi-vibrator is a rectangular wave generating circuit. The circuit does not require any external trigger to change the output &amp; hence the name free running. The time during which the output is either high or low is determined by the two resistors &amp; capacitors which are connected externally.</p> <p>Fig.1 shows the 555 timer connected as an astable multivibrator. To understand the circuit operations consider the internal block diagram of the 555 timer.</p> <p>Initially when output is high, capacitor C starts towards <math>V_{CC}</math> through <math>R_A</math> and <math>R_B</math>. However, as soon as voltage across the capacitor equals <math>\frac{2}{3} V_{CC}</math>, comparator -1 triggers the flip-flop and the o/p switches low. Now capacitor C starts discharging through <math>R_B</math> and transistor <math>Q_1</math>. When the voltage across C equals <math>\frac{1}{3} V_{CC}</math>, comparators output triggers the flip-flop and output goes to high. Then the cycle repeats the output voltage and</p>				



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capacitor voltage waveform as shown in Fig. 2.



The time during which the capacitor charges from  $\frac{1}{3} V_{CC}$  to  $\frac{2}{3} V_{CC}$  is equal to the time the output is high & is given by

$$t_C = 0.69(R_A + R_B) C$$

The time during which, the capacitor discharges from  $\frac{2}{3} V_{CC}$  to  $\frac{1}{3} V_{CC}$  is equal to the time the output is low & is given by

$$t_d = 0.69(R_B) C$$

$$\therefore \text{Total period } T = t_C + t_d$$

$$= 0.69(R_A + R_B) C$$

$$\text{Hence frequency of oscillator is } f_o = \frac{1}{T} = \frac{1.45}{(R_A + 2R_B) C}$$

The duty cycle is the ratio of time  $t_C$  during which the output is high to the total period T.

$$\% \text{ duty cycle} = \frac{t_C}{T} * 100$$

$$= \frac{R_A + R_B}{R_A + 2R_B} * 100\%$$

As seen from above equation, astable multi-vibrator will not produce square wave unless the resistance  $R_A = 0$ . With  $R_A = 0$ , pin 7 is directly connected to



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		<p><math>V_{CC}</math> &amp; when capacitor discharge through <math>R_B</math>, an extra current is supplied to <math>Q_1</math> by <math>V_{CC}</math> through short between 7 &amp; <math>V_{CC}</math> which damages <math>Q_1</math> &amp; hence timer.</p> <p>Without reducing <math>R_A</math> to <math>0\Omega</math>, the astable multi-vibrator can produce square wave output simply by connecting diode across <math>R_B</math> as shown in Fig.3.</p> <p><b>2. Monostable multivibrator:</b></p> <p>A monostable multi-vibrator is often called as one shot multi-vibrator. It is a pulse generating circuit in which the duration of the pulse is determined by the RC network connected externally. When an external trigger pulse is applied, the output is forced to go high. The time the output remains high is determined by the external RC network connected to the timer. At the end of time interval, the output automatically reverse back to its logic low stable states. The output stays low until the trigger pulse is again applied. Then the cycle repeats. The monostable circuit has only one stable state &amp; hence the name monostable.</p> <p><b>Circuit operation:</b></p> <p>The circuit is shown in Fig. 5. Initially, when the output is low, transistor <math>Q_1</math> is on &amp; capacitor C is shorted to ground. However, upon the application of negative trigger pulse to pin 2, transistor <math>Q_1</math> is turned off, which releases the short circuit across the external capacitor C &amp; drives the output high. The capacitor C now starts charging up towards <math>V_{CC}</math> through <math>R_A</math>.</p> <p>However, when the voltage across the capacitor equals <math>\frac{2}{3} V_{CC}</math>, comparator 1 output switch from low to high which in turn drives the output to its low state via the flip-flop. The output of flip-flop turns <math>Q_1</math> ON &amp; hence capacitor C rapidly discharges through the transistor. The output remains low until a trigger pulse is applied again. The time during which the output remains high is given by</p> $t_p = 0.69 R_A C$
6	Procedure, Program, Activity, Algorithm, Pseudo Code	<p><b>Procedure:</b></p> <p><b>Astable multi-vibrator:</b></p> <p><b>i. D&gt;50%</b></p> <p>1. Verify the components for its working.</p>



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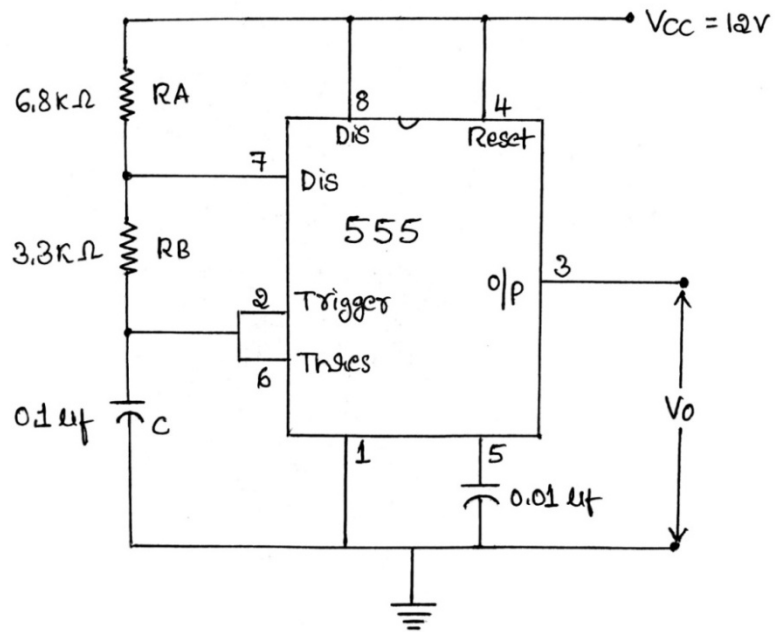
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		<ol style="list-style-type: none"> <li>2. Connect the Astable multi-vibrator circuit as shown in Fig.1.</li> <li>3. Switch the DC power supply (<math>V_{CC} = 12V</math>) and observe the output waveform. On CRO at pin 3.</li> <li>4. Measure the output pulse amplitude.</li> <li>5. Observe capacitor voltage waveform at pin 6 &amp; measure the maximum &amp; minimum levels (<math>V_{UT}</math> &amp; <math>V_{LT}</math>).</li> <li>6. Calculate duty cycle 'd' &amp; output frequency f &amp; verify with theoretical values.</li> </ol> <p><b>ii) For <math>D \leq 50\%</math></b></p> <ol style="list-style-type: none"> <li>1. Connect the circuit as shown in Fig. 3.</li> <li>2. Set <math>R_A = R_B = R</math> [For 50% duty cycle].</li> <li>3. Select suitable value for C.</li> <li>4. Measure output voltage &amp; note capacitor voltage on CRO &amp; measure <math>V_{UT}</math>, <math>V_{LT}</math>.</li> <li>5. Calculate d &amp; output frequency f.</li> </ol> <p><b>Monostable multi-vibrator:</b></p> <ol style="list-style-type: none"> <li>1. Check the components for its working.</li> <li>2. Connect the circuit as shown in Fig. 5.</li> <li>3. Switch power supply ON &amp; apply periodic input trigger pulse [negative going trigger] at pin 2 using pulse generator (or signal generator).</li> </ol> <ol style="list-style-type: none"> <li>4. Adjust input frequency of pulse generator to 80Hz &amp; adjust the input pulse amplitude to 12V.</li> <li>5. Observe timer output at pin 3 on CRO &amp; calculate f.</li> </ol> <p><b>NOTE:</b> If <math>T = 1</math> ms, then <math>f = 1</math>KHz, Set input freq <math>\approx 1</math>KHz say 800/500Hz  If <math>T = 10</math> MS, then <math>f = 100</math>Hz, Set input freq <math>\approx 100</math>Hz say 80Hz</p>
7	Block, Circuit, Model Diagram, Reaction Equation, Expected Graph	<p><b>1. Astable multivibrator</b></p> <p><b>Circuit diagram:</b></p> <p><b>I. Un- symmetrical Astable multivibrator:</b></p>

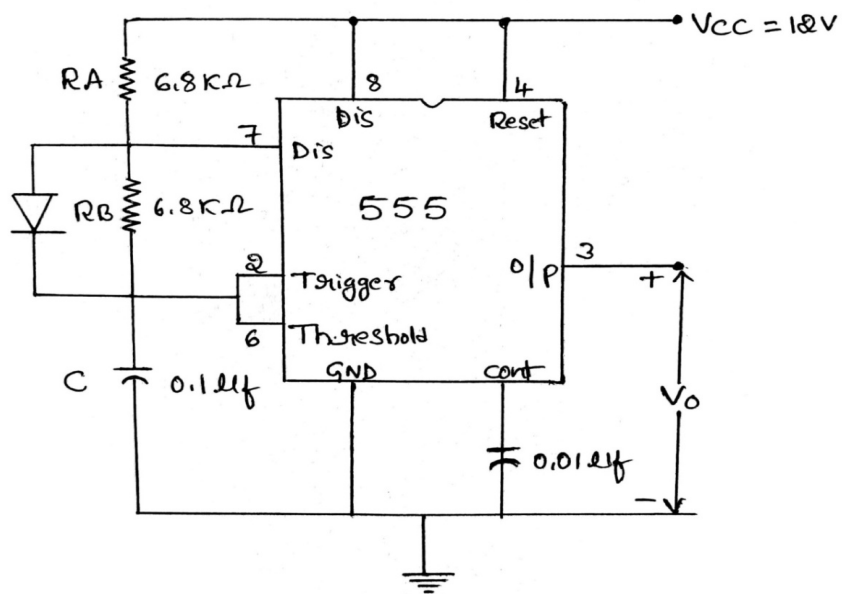


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**II. Symmetrical Astable multivibrator:**



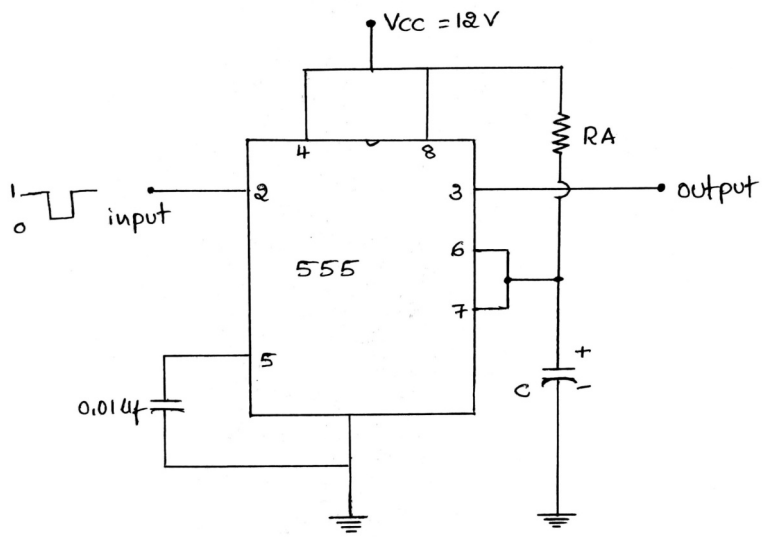
**Monostable multivibrator:**

**Circuit diagram:**



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8 Observation Table, Look-up Table, Output

$V_{UT}$ (V)	$V_{LT}$ (V)	$T_H$ (mS)	$T_L$ (mS)	T(mS)
Theoretical = $\frac{2}{3}V_{CC}$	Theoretical = $\frac{1}{3}V_{CC}$	Theoretical	Theoretical	Theoretical
Practical	Practical	Practical	Practical	Practical

monostable multivibrator

T(ms)		$\frac{2}{3} V_{CC}$ (V)	
Theoretical	Practical	Theoretical	Practical

9 Sample Calculations

**Design:**

Design Astable multi-vibrator to produce 1KHz square wave with duty cycle of 75%.

**Solution:**

Given duty cycle,  $D=0.75$ ,  $f = 1 \text{ KHz}$

$$\text{Time period } T = \frac{1}{f} = \frac{1}{1 \text{ KHz}} = 1 \text{ ms}$$

$$\text{But } T = T_H + T_L$$





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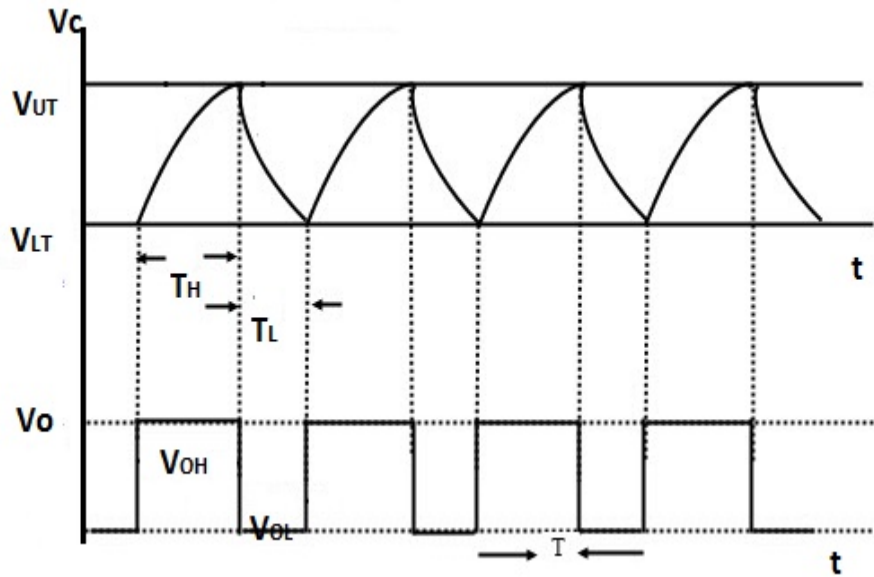
$\therefore 1\text{ms} = T_H + T_L \dots\dots\dots(1)$   
 Duty cycle,  $D = \frac{T_H}{T}$   
 $0.75 = \frac{T_H}{T} \dots\dots\dots T_H = 0.75$   
 $T_H = 0.75(1\text{ms})$   
 $T_H = 0.75\text{ms} \dots\dots\dots(2)$   
 Substitute Eq. (2) in Eq. (1) gives  
 $\therefore T_L = T - T_H$   
 $= 1\text{m} - 0.75\text{m}$   
 $T_L = 0.25\text{ms}$   
 But  $T_L = 0.69 R_B C$   
 Let  $C = 0.1\mu\text{F}$   
 $\therefore T_L = 0.69 * R_B * 0.1\mu$   
 $\therefore R_B = \frac{T_L}{0.69 * 0.1\mu} = \frac{0.25\text{m}}{0.69 * 0.1\mu} = 3.6\text{K}\Omega$   
 $R_B = 3.6\text{K}\Omega$   
 But  $T_H = 0.69(R_A + R_B) C$   
 $0.75\mu\text{f} = 0.69(R_A + R_B) 0.1\mu$   
 $R_A + R_B = 10.82\text{K}$   
 $\therefore R_A = 10.82\text{K} - R_B = 7.2\text{K}\Omega$   
 $\therefore$  Select  $R_A = 6.8\text{K}\Omega$  &  $R_B = 3.3\text{K}\Omega$   
  
**Design monostable multi-vibrator having time delay  $t_p = 10\text{ms}$**   
**Solution:**  
 Given  $t_p = 10\text{ms}$   
 Choose  $C = 0.1\mu\text{f}$   
 $\therefore R_A = \frac{t_p}{1.1C} = \frac{10 * 10^{-3}}{1.1 * 0.1 * 10^{-6}} = 90\text{K}\Omega$   
 $\therefore$  Select  $R_A = 100\text{K}\Omega$

10 Graphs, Outputs



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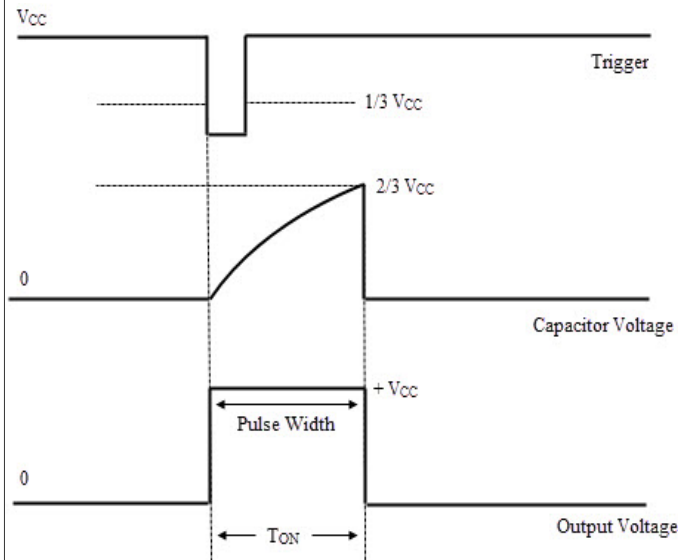
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$$V_{UT} = \frac{2}{3} V_{CC}, \quad V_{LT} = \frac{1}{3} V_{CC}$$

Note: For symmetrical astable multi-vibrator  $V_{OH} = V_{OL}$

Monostable multi-vibrator



11	Results & Analysis	<ol style="list-style-type: none"> <li>1. Unsymmetrical Astable multivibrator was designed and verified with % duty cycle =.....</li> <li>2. symmetrical Astable multivibrator was designed and verified</li> <li>3. Operation of monostable multi-vibrator is verified.</li> </ol>
12	Application Areas	<p><b>Astable multivibrator:</b></p> <ol style="list-style-type: none"> <li>1. Square wave oscillator.</li> </ol>



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		2 Free running ramp generator. <b>Monostable multivibrator:</b> 1. Frequency divider. 2. Pulse stretcher.
13	Remarks	
14	Faculty Signature with Date	

### Experiment 09 : RC PHASE SHIFT OSCILLATOR AND HARTLEY OSCILLATOR

-	<b>Experiment No.:</b>	9	<b>Marks</b>	10	<b>Date Planned</b>		<b>Date Conducted</b>																																																					
1	Title	RC Phase shift oscillator and Hartley oscillator																																																										
2	Course Outcomes	Simulate and analyze analog circuits that uses transistor and ICs for different electronic applications.																																																										
3	Aim	To design and test the following tuned oscillator circuits for the given frequency. (a) RC Phase Shift Oscillator using BJT. (b) Hartley Oscillator using BJT.																																																										
4	Material Equipment Required	<table border="1"> <thead> <tr> <th>Sl.No.</th> <th>Particulars</th> <th>Specification</th> <th>Quantity</th> </tr> </thead> <tbody> <tr> <td>1</td> <td>Transistor</td> <td>SL100</td> <td></td> </tr> <tr> <td rowspan="5">2</td> <td rowspan="5">Resistor</td> <td>1.2K<math>\Omega</math></td> <td>1</td> </tr> <tr> <td>1.8K <math>\Omega</math></td> <td>2</td> </tr> <tr> <td>4.7K <math>\Omega</math></td> <td>3</td> </tr> <tr> <td>12K <math>\Omega</math></td> <td>1</td> </tr> <tr> <td>62K <math>\Omega</math></td> <td>1</td> </tr> <tr> <td>3</td> <td>DRB</td> <td>10K <math>\Omega</math></td> <td>1</td> </tr> <tr> <td rowspan="3">4</td> <td rowspan="3">Capacitor</td> <td>0.47<math>\mu</math>f</td> <td>2</td> </tr> <tr> <td>0.01<math>\mu</math>f</td> <td>3</td> </tr> <tr> <td>4.7<math>\mu</math>f</td> <td>1</td> </tr> <tr> <td>5</td> <td>DCB</td> <td></td> <td>1</td> </tr> <tr> <td>6</td> <td>VRPS</td> <td>0 – 32 V</td> <td>2</td> </tr> <tr> <td>7</td> <td>CRO</td> <td>20 MHz, Dual Channel</td> <td>1</td> </tr> <tr> <td>8</td> <td>CRO Probes</td> <td></td> <td>3</td> </tr> <tr> <td>9</td> <td>Digital Multimeter</td> <td></td> <td>1</td> </tr> </tbody> </table>							Sl.No.	Particulars	Specification	Quantity	1	Transistor	SL100		2	Resistor	1.2K $\Omega$	1	1.8K $\Omega$	2	4.7K $\Omega$	3	12K $\Omega$	1	62K $\Omega$	1	3	DRB	10K $\Omega$	1	4	Capacitor	0.47 $\mu$ f	2	0.01 $\mu$ f	3	4.7 $\mu$ f	1	5	DCB		1	6	VRPS	0 – 32 V	2	7	CRO	20 MHz, Dual Channel	1	8	CRO Probes		3	9	Digital Multimeter		1
Sl.No.	Particulars	Specification	Quantity																																																									
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### Hartley Oscillator using BJT.

Sl.No.	Particulars	Specification	Quantity
1	BJT		2
2	capacitors Resistor	0.1 $\mu$ f	2
		4.7 $\mu$ f	1
		8.84nF	1
		330 $\Omega$	1
		820 $\Omega$	1
3		2M $\Omega$	1
4	Inductors CRO	240 $\mu$ H	1
		100 $\mu$ H	1
		Dual Channel	1

5 Theory, Formula, Principle, Concept

### RC PHASE SHIFT:

An oscillator is an electronic circuit for generating an AC signal voltage with a DC supply as the

Only input requirement. The frequency of the generated signal is decided by the circuit elements used. An oscillator requires an amplifier, a frequency selective network and a positive feedback from the output to the input. The Barkhausen criterion for sustained oscillation is  $A\beta = 1$  where A is the gain of the amplifier and  $\beta$  is the feedback factor (gain). The unity gain means signal is in phase. ( If the signal is 180° out of phase and gain will be -1). RC-Phase shift Oscillator has a CE amplifier followed by three sections of RC phase shift feed-back Networks. The output of the last stage is return to the input of the amplifier. The values of R and C are chosen such that the phase shift of each RC section is 60°. Thus The RC ladder network produces a total phase shift of 180° between its input and output voltage for the given frequency. Since CE Amplifier produces 180° phase shift. The total phase shift from the base of the transistor around the circuit and back to the base will be exactly 360° or 0°. This satisfies the Barkhausen condition for sustaining oscillations and total loop gain of this circuit is greater than or equal to 1, this condition used to generate the sinusoidal oscillations

### HARTLEY:

An oscillator is an electronic circuit for generating an AC signal voltage with a DC supply as the

Only input requirement. The frequency of the generated signal is decided by the circuit elements used. An oscillator requires an amplifier, a frequency selective network and a positive feedback from the output to the input. The Barkhausen criterion for sustained oscillation is  $A\beta = 1$  where A is the gain of the amplifier and  $\beta$  is the feedback factor (gain). The unity gain means signal is in phase The Hartley oscillator is an electronic oscillator circuit in which the oscillation frequency is determined by a tuned circuit consisting of capacitors and inductors, that is, an LC oscillator. The Hartley oscillator is distinguished by a tank circuit consisting of two series-connected coils (or, often, a tapped coil) in parallel with a capacitor, with an amplifier between the relatively high impedance across the entire LC tank and the



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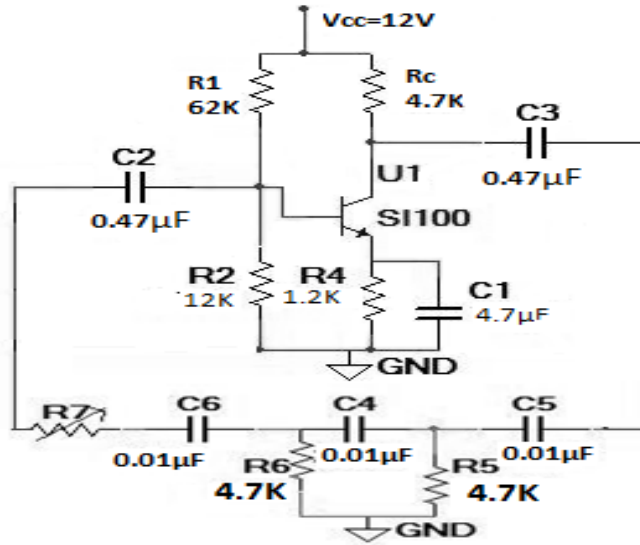
		<p>relatively low voltage/high current point between the coils. The Hartley oscillator is the dual of the Colpitts oscillator which uses a voltage divider made of two capacitors rather than two inductors. Although there is no requirement for there to be mutual coupling between the two coil segments, the circuit is usually implemented using a tapped coil, with the feedback taken from the tap, as shown here. The optimal tapping point (or ratio of coil inductances) depends on the amplifying device used, which may be a bipolar junction transistor.</p>
6	<p>Procedure, Program, Activity, Algorithm, Pseudo Code</p>	<ol style="list-style-type: none"> <li>1. Rig up the circuit as shown in figure</li> <li>2. place all the required components from multi-sim library</li> <li>3. click on run button and observe the out put</li> </ol> <p><b>To plot frequency response</b></p> <ol style="list-style-type: none"> <li>1. Place the components in Multisim.</li> <li>2. Rig up the circuit as shown in circuit diagram.</li> <li>3. Click on the run button and Double click on Oscilloscope.</li> <li>4. Observe the output waveforms on Oscilloscope.</li> <li>5. Measure frequency of the output signal, compare it with theoretical frequency.</li> </ol> <p>HARTLEY:</p> <p><b>6. To plot frequency response</b></p> <ol style="list-style-type: none"> <li>7. Place the components in Multisim.</li> <li>8. Rig up the circuit as shown in circuit diagram.</li> <li>9. Click on the run button and Double click on Oscilloscope.</li> <li>10. Observe the output waveforms on Oscilloscope.</li> <li>11. Measure frequency of the output signal, compare it with theoretical frequency.</li> </ol>



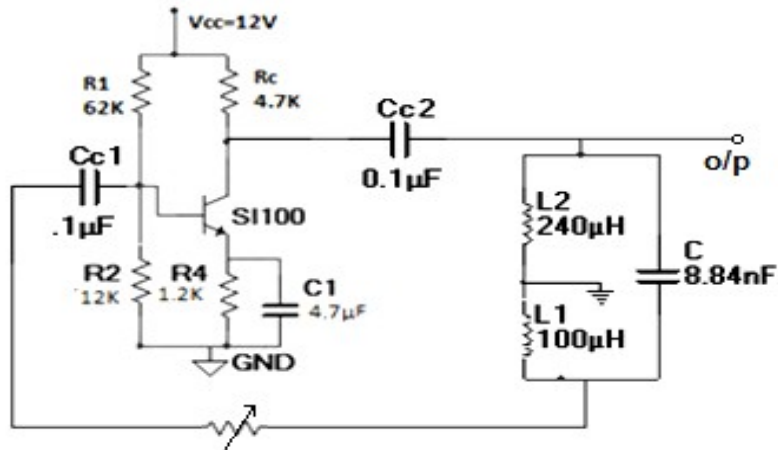
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7 Block, Circuit, Model Diagram, Reaction Equation, Expected Graph



HARTLEY



8 Observation Table, Look-up Table, Output

9 Sample Calculations

10 Graphs, Outputs



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11	Results & Analysis	The RC Phase Shift Oscillator and Hartley simulated and output is verified
12	Application Areas	RC phase shift oscillator is used in musical instrument, GPS unit and in voice synthesis Hartley oscillator are used in radio receivers
13	Remarks	
14	Faculty Signature with Date	

### Experiment 10 : NARROW BAND-PASS FILTER AND NARROW BAND-REJECT FILTER

-	Experiment No.:	10	Marks	10	Date Planned	Date Conducted
1	Title	Narrow Band-pass Filter and Narrow band-reject filter				
2	Course Outcomes	Simulate and analyze analog circuits that uses transistor and ICs for different electronic applications.				
3	Aim	To simulate and analyze the Narrow Band-pass Filter and Narrow band-reject filter				
4	Material Equipment Required	/				
5	Theory, Formula, Principle, Concept	<p>A narrow bandpass filter employing multiple feedback is depicted in figure. This filter employs only one op-amp, as shown in the figure. In comparison to all the filters discussed so far, this filter has some unique features that are given below.</p> <p><b>1. It has two feedback paths, and this is the reason that it is called a multiple-feedback filter.</b></p> <p><b>2. The op-amp is used in the inverting mode.</b></p>				



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The frequency response of a narrow bandpass filter is shown in fig(b).

Generally, the narrow bandpass filter is designed for specific values of centre frequency  $f_c$  and  $Q$  or  $f_c$  and  $BW$ . The circuit components are determined from the following relationships. For simplification of design calculations each of  **$C_1$  and  $C_2$  may be taken equal to  $C$** .

$$R_1 = Q/2\pi f_c CA_f$$

$$R_2 = Q/2\pi f_c C(2Q^2 - A_f)$$

$$\text{and } R_3 = Q / \pi f_c C$$

where  $A_f$  is the gain at centre frequency and is given as

$$A_f = R_3 / 2R_1$$

The gain  $A_f$  however must satisfy the condition  $A_f < 2 Q^2$ .

The centre frequency  $f_c$  of the multiple feedback filter can be changed to a new frequency  $f'_c$  without changing, the gain or bandwidth. This is achieved simply by changing  $R_2$  to  $R'_2$  so that

$$R'_2 = R_2 [f_c/f'_c]^2$$

#### **BAND REJECT FILTER:**

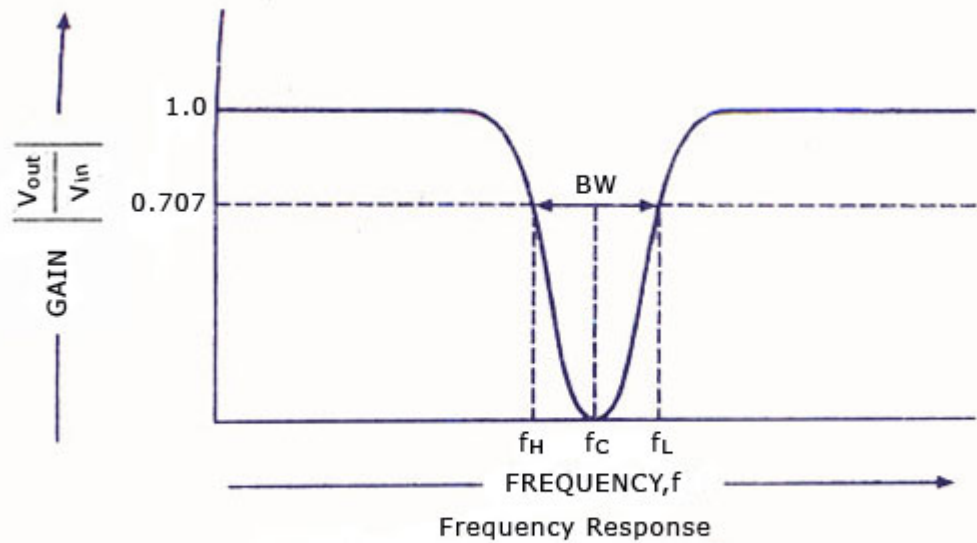
**This is also called a notch filter.** It is commonly used for attenuation of a single frequency such as 60 Hz power line frequency hum. The most widely used notch filter is the twin-T network illustrated in fig. (a). This is a passive filter composed of two T-shaped networks. One T-network is made up of two resistors and a capacitor, while the other is made of two capacitors and a resistor. One drawback of above notch filter (passive twin-T network) is that it has relatively low figure of merit  $Q$ . However,  $Q$  of the network can be increased significantly if it is used with the voltage follower, as illustrated in fig. (a). Here the output of the voltage follower is supplied back to the junction of  $R/2$  and  $2C$ . The frequency response of the active notch filter is shown in fig (b).





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Twin T Active Notch Filter

www.CircuitsToday.com

Notch filters are most commonly used in communications and biomedical instruments for eliminating the undesired frequencies.

A mathematical analysis of this circuit shows that it acts as a lead-lag circuit with a phase angle, shown in fig. (b). Again, there is a frequency  $f_c$  at which the phase shift is equal to  $0^\circ$ . In fig. (c), the voltage gain is equal to 1 at low and high frequencies. In between, there is a frequency  $f_c$  at which voltage gain drops to zero. Thus such a filter notches out, or blocks frequencies near  $f_c$ . The frequency at which maximum attenuation occurs is called the notch-out frequency given by

$$f_n = F_c = 2\pi RC$$

Notice that two upper capacitors are  $C$  while the capacitor in the centre of the network is  $2C$ . Similarly, the two lower resistors are  $R$  but the resistor in the centre of the network is  $1/2 R$ . This relationship must always be maintained.

Let us consider the narrow band notch filter circuit. We know that the notch filter is used to eliminate single frequency. Thus let us consider the frequency to eliminate be 120 Hz. The capacitor value  $C = 0.33 \mu F$ .

By using the centre frequency  $f_c = 1/(2\pi RC)$

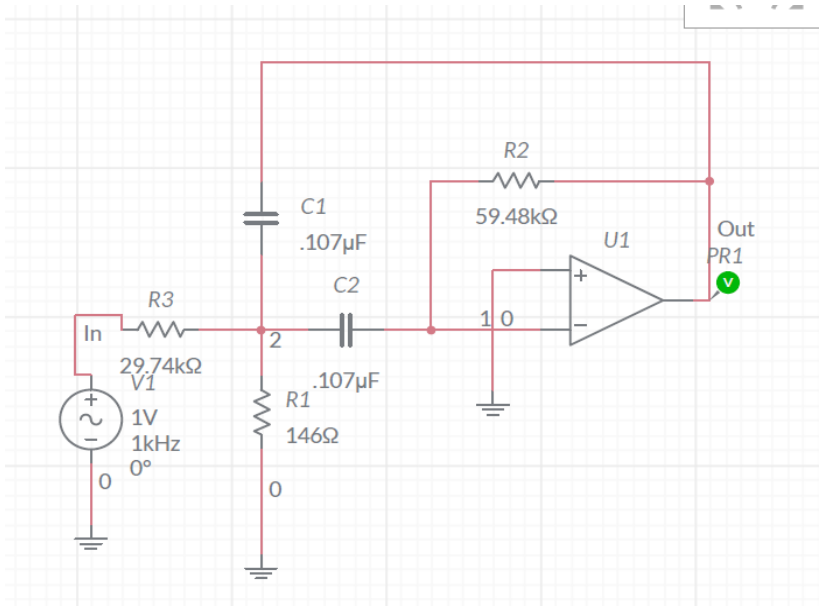
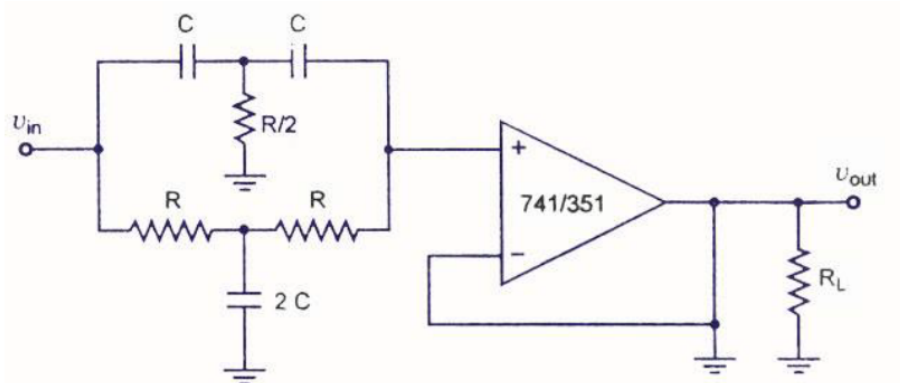
$$R = 1/(2\pi f_c C) = 1 / (2\pi \times 120 \times 0.33 \times 10^{-6}) = 4 \text{ k}\Omega$$

Thus, in order design the notch filter to eliminate 120 Hz frequency we have to take two parallel resistors with  $4 \text{ k}\Omega$  each and the two capacitors in parallel with  $0.33 \mu F$  each.



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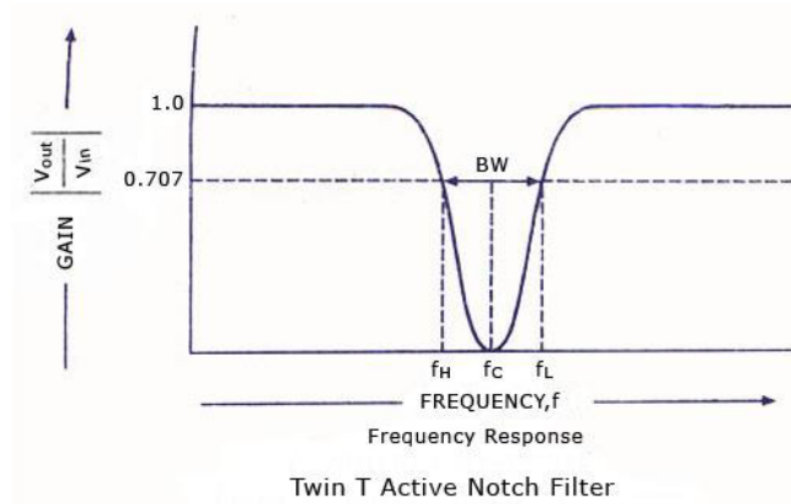
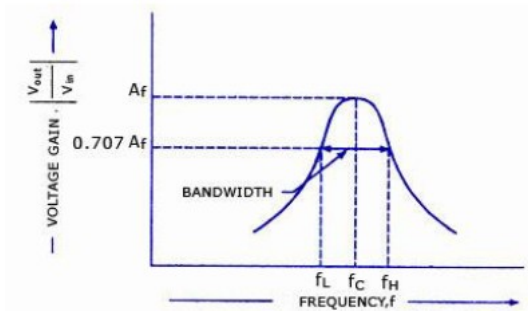
6	Procedure, Program, Activity, Algorithm, Pseudo Code	<ol style="list-style-type: none"> <li>1. Rig up the circuit as shown in figure</li> <li>2. place all the required components from multi-sim library</li> <li>3. click on run button and observe the out put</li> </ol>
7	Block, Model, Circuit, Diagram, Reaction Equation, Expected Graph	 <p>Narrow Band-Stop Filter.</p>  <p>Circuit Diagram</p>
8	Observation Table, Look-up Table, Output	
9	Sample Calculations	



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10 Graphs, Outputs



11	Results & Analysis	The band-pass and band-reject filters are simulated and output is verified
12	Application Areas	Narrow band-pass used in wireless transceiver Band reject filter is used in speaker systems , telephone technology
13	Remarks	
14	Faculty Signature with Date	

**Experiment 11 :PRECISION HALF AND FULL WAVE RECTIFIER**

-	Experiment No.:	11	Marks	10	Date Planned		Date Conducted	
1	Title	<b>Precision Half and full wave rectifier</b>						
2	Course Outcomes	Simulate and analyze analog circuits that uses transistor and ICs for different electronic applications.						
3	Aim	<b>To simulate and analyze Precision Half and full wave rectifier</b>						
4	Material Equipment Required	/						
5	Theory, Formula,	When forward biased voltage is less than 0.7V, then diode is not conducting. In						

Dept EC

Prepared by

Checked by

Approved



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	Principle, Concept	<p>case of normal power rectifier input applied is much larger than 0.7V. So diode is not operated. Therefore Op-amp is used to help diode to conduct. The precision rectifiers are classified in two categories.</p> <p><b>1. Precision Half wave rectifier</b></p> <p><b>2. Precision Full wave Rectifier</b></p> <p><b>1. Precision Half wave rectifier (HWR) :</b></p> <p>In HWR, the diode conducts in one of the half cycles of applied ac input signal. Because of this we can classify HWR as positive PHWR (output is positive) and negative PHWR (output is negative).</p> <p>In positive half cycle of applied ac input signal output of op-amp is negative, so diode D1 is forward biased and D2 is reversed biased. The output of op-amp is virtually shorted to ground and output voltage is zero.</p> <p>In negative half cycle of applied ac input signal output of op-amp is positive, so diode D2 is forward biased and D1 is reversed biased.</p> <p>Non-saturated types of precision half wave rectifiers are suitable for high frequency applications. In HWR, the diode conducts in one of the half cycles of applied ac input signal.</p> <p>Design: In positive half cycle of applied ac input signal output of op-amp is negative, so diode D1 is forward biased and D2 is reversed biased. The output of op-amp is virtually shorted to ground and prevented going into saturation. Thus output voltage is zero.</p> <p><math>\therefore V_o = 0 \text{ V}</math></p> <p>In negative half cycle of applied ac input signal output of op-amp is positive, so diode D2 is forward biased and D1 is reversed biased. The circuit now works as an inverting amplifier with gain of <math>(-R_f/R_1)</math></p> <p><math>V_o = V_{in} \times A</math></p> <p>But in negative half cycle input magnitude is negative therefore we get,</p> <p style="text-align: right;"><math>V_o = (-V_{in})[-R_f/R_1]</math></p> <p><math>\therefore V_o = R_f/R_1(V_{in})</math></p> <p>Thus in negative half cycle output is positive with a gain of <math>(R_f/R_1)</math>.</p> <p><b>2. Precision Full wave Rectifier:</b></p> <p>In PFWR, for both the half cycles output is produced &amp; in one direction only. In positive half cycle of applied ac input signal, output of first op-amp (A1) is Negative. Therefore diode D2 is forward biased &amp; diode D1 is reverse biased.</p>
6	Procedure, Program, Activity, Algorithm, Pseudo Code	
7	Block, Circuit, Model Diagram, Reaction Equation, Expected Graph	Half-wave:

Dept EC

Prepared by

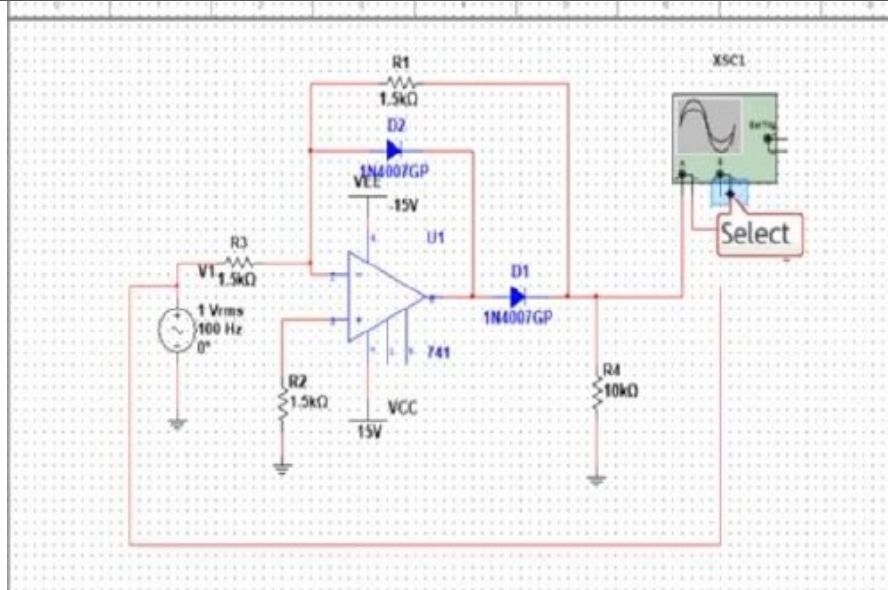
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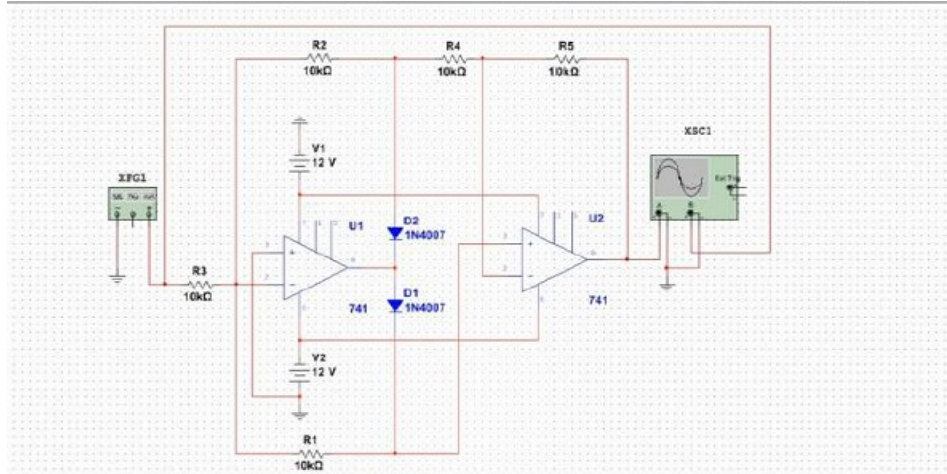


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full wave:

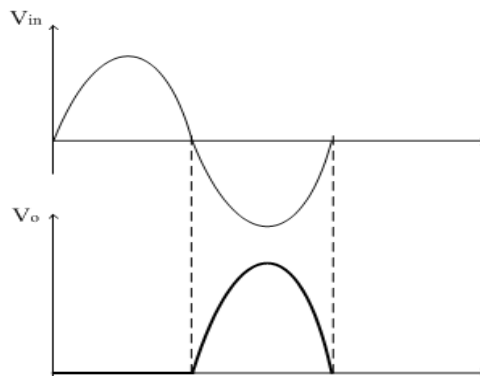


8 Observation Table, Look-up Table, Output

9 Sample Calculations

10 Graphs, Outputs

Half wave :





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		<p>full wave:</p>
11	Results & Analysis	The Precision Half and full wave rectifier simulated and output is verified
12	Application Areas	Used in instrumentation application
13	Remarks	
14	Faculty Signature with Date	

### Experiment 12 : Monostable and Astable multivibrator

-	Experiment No.:	12	Marks	10	Date Planned	Date Conducted
1	Title	<b>Monostable and Astable multivibrator</b>				
2	Course Outcomes	Simulate and analyze analog circuits that uses transistor and ICs for different electronic applications.				
3	Aim	To simulate and analyze Monostable and Astable multivibrator				
4	Material Equipment Required	/				
5	Theory, Formula, Principle, Concept	<p><b>Astable multivibrator:</b></p> <p>The basic 555 timer based Astable multivibrator circuit is depicted in fig 1. Initially, capacitor C is fully discharged, which forces the output to go to the HIGH state. An open discharge transistor allows capacitor C to charge from +V<sub>cc</sub> through resistors R1 and R2. When the voltage across C exceeds <math>+2V_{cc} / 3</math>, the output enters the LOW state and the discharge transistor is switched ON at the same time. Capacitor C starts to discharge through R2 and the discharge transistor inside the IC.</p> <p>When the voltage across C falls below <math>+ V_{cc} / 3</math>, the output enters the HIGH state. The charge and discharge cycles repeat and the circuit behaves as a free running multivibrator. Terminal-4 of the IC is the RESET terminal. Usually, it is connected to +V<sub>cc</sub>. If the voltage at this terminal is driven below 0.4, the output is forced to the LOW state overriding command pulses at terminal-2 of the IC. HIGH</p>				



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state and LOW state time periods are governed by the charge (+ Vcc /3 to +2Vcc /3) and discharge (+2Vcc /3 to +Vcc /3) timings. These are given by (1) and (2) respectively.

HIGH state time period,  $T_{HIGH} = 0.69 \times (R1+R2) \times C$

LOW state time period,  $T_{LOW} = 0.69 \times R2 \times C$

6 Procedure, Program, Activity, Algorithm, Pseudo Code

1. Rig up the circuit as shown in figure
2. place all the required components from multi-sim library
3. click on run button and observe the out put

7 Block, Circuit, Model Diagram, Reaction Equation, Expected Graph

**Astable multivibrator:**

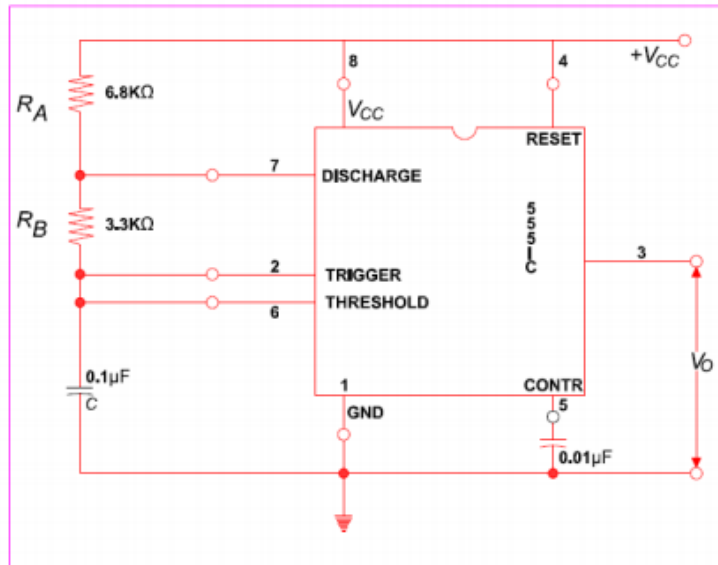
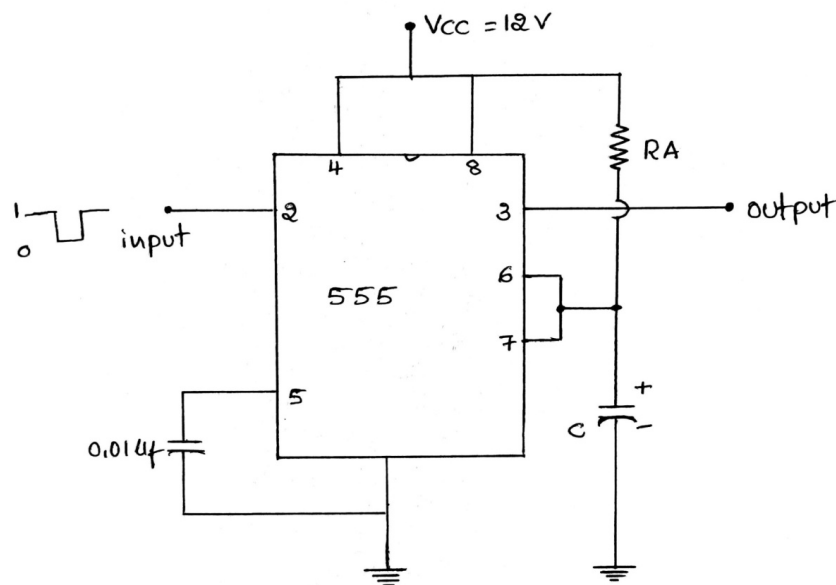


Fig. 3 Astable multivibrator circuit schematic

monostable multi-vibrator:



8 Observation Table,





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Look-up Table, Output	
9 Sample Calculations	
10 Graphs, Outputs	<p>Astable multi-vibrator</p> <p>monostable multi-vibrator</p>
11 Results & Analysis	The Astable and manostable multi-vibrator is simulated and output is verified
12 Application Areas	<p><b>Astable multivibrator:</b></p> <ul style="list-style-type: none"> <li>2. Square wave oscillator.</li> <li>2 Free running ramp generator.</li> </ul>





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		<b>Monostable multivibrator:</b> 2. Frequency divider. 2. Pulse stretcher.
13	Remarks	
14	Faculty Signature with Date	